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Example of Laplace node analysis of an RLLCC circuit.

## Classical RLLCC Node Analysis Example

First write KCL node equations:

$$C_1 p(v_1 - v_a) + \frac{1}{R}(v_1 - v_2) + \frac{1}{L_1 p}v_1 = 0$$

Node  $v_1$

$$\frac{1}{L_2 p}(v_2 - v_a) + \frac{1}{R}(v_2 - v_1) + C_2 p v_2 = 0$$

Node  $v_2$

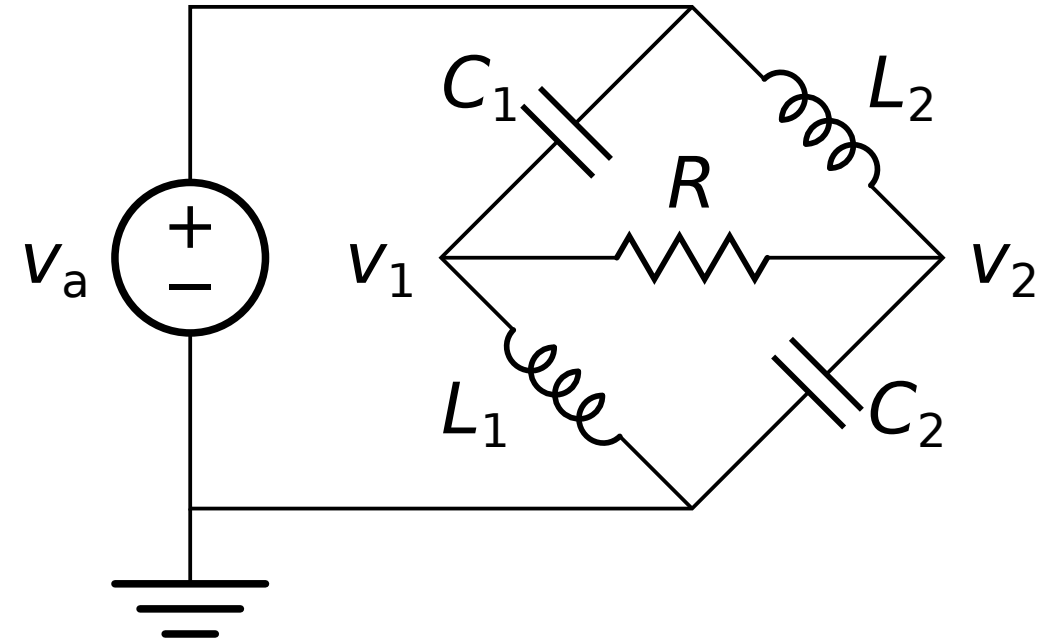
For put them in normal form:

$$\left(C_1 p + \frac{1}{R} + \frac{1}{L_1 p}\right)v_1 + \left(-\frac{1}{R}\right)v_2 = C_1 p v_a$$

Node  $v_1$

$$\left(-\frac{1}{R}\right)v_1 + \left(C_2 p + \frac{1}{R} + \frac{1}{L_2 p}\right)v_2 = \frac{1}{L_2 p}v_a$$

Node  $v_2$



Solving by Hand:

$$\text{Raw Node } v_1 : C_1 p(v_1 - v_a) + \frac{1}{R}(v_1 - v_2) + \frac{1}{L_1 p}v_1 = 0 \quad \text{Node } v_2 : \frac{1}{L_2 p}(v_2 - v_a) + \frac{1}{R}(v_2 - v_1) + C_2 p v_2 = 0$$

Solving for  $v_2$  in the  $v_1$  equation yields:  $v_2 = RC_1 p v_1 - RC_1 p v_a + v_1 + \frac{R}{L_1 p}v_1$ .

Substituting this  $v_2$  into the  $v_2$  equation:

$$(RC_1 C_2) p^2 v_1 + (C_1 + C_2) p v_1 + \left( \frac{RC_1}{L_2} + \frac{RC_2}{L_1} \right) v_1 + \left( \frac{1}{L_1} + \frac{1}{L_2} \right) \frac{1}{p} v_1 + \frac{R}{L_1 L_2} \frac{1}{p^2} v_1 = \left( RC_1 C_2 p^2 + C_1 p + \frac{RC_1}{L_2} + \frac{1}{L_2 p} \right) v_a$$

To make things easier  $C_1 \rightarrow C_2 \rightarrow C$  and  $L_1 \rightarrow L_2 \rightarrow L$ :

$$RC^2 p^2 v_1 + 2C p v_1 + \frac{2RC}{L} v_1 + \frac{2}{Lp} v_1 + \frac{R}{L^2 p^2} v_1 = \left( RC^2 p^2 + Cp + \frac{RC}{L} + \frac{1}{Lp} \right) v_a$$

## Classical RLLCC Node Analysis Example

Next, put in normal form. (Here multiply by  $\frac{p^2}{RC^2}$ ):

$$\left(p^4 + \frac{2}{RC}p^3 + \frac{2}{LC}p^2 + \frac{2}{RLC^2}p + \frac{1}{(LC)^2}\right)v_1 = \left(p^4 + \frac{1}{RC}p^3 + \frac{1}{LC}p^2 + \frac{1}{RLC^2}p\right)v_a$$

The characteristic equation:

$$p^4 + \frac{2}{RC}p^3 + \frac{2}{LC}p^2 + \frac{2}{RLC^2}p + \frac{1}{(LC)^2} = 0$$

And its solution:

$$\left(p + \sqrt{-\frac{1}{LC}}\right) \left(p - \sqrt{-\frac{1}{LC}}\right) \left(p + \frac{1}{RC} + \sqrt{\frac{1}{(RC)^2} - \frac{1}{LC}}\right) \left(p + \frac{1}{RC} - \sqrt{\frac{1}{(RC)^2} - \frac{1}{LC}}\right) = 0$$

The complementary (transient) solution (assuming all four roots are distinct):

$$\begin{aligned} v_1(t) &= K_1 e^{a_1 t} + K_2 e^{a_2 t} + K_3 e^{a_3 t} + K_4 e^{a_4 t} \\ &= K_1 e^{-jt \frac{1}{\sqrt{LC}}} + K_2 e^{jt \frac{1}{\sqrt{LC}}} + K_3 e^{-t \frac{1}{RC}} e^{t((RC)^{-2} - (LC)^{-1})^{1/2}} + K_4 e^{-t \frac{1}{RC}} e^{-t((RC)^{-2} - (LC)^{-1})^{1/2}} \end{aligned}$$

assuming that  $K_1 = K_2$  and  $K_1$  is real

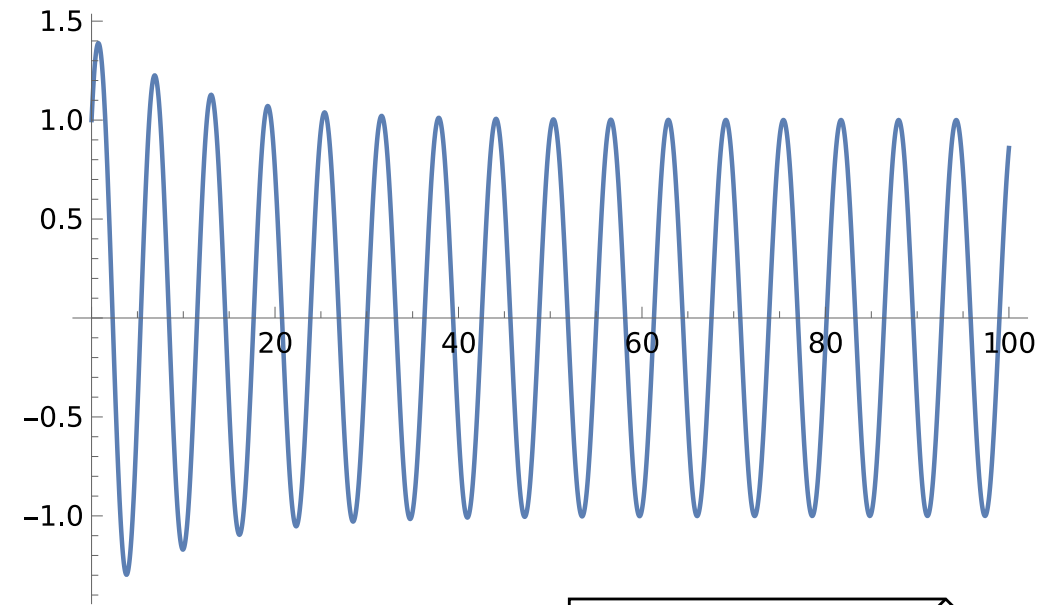
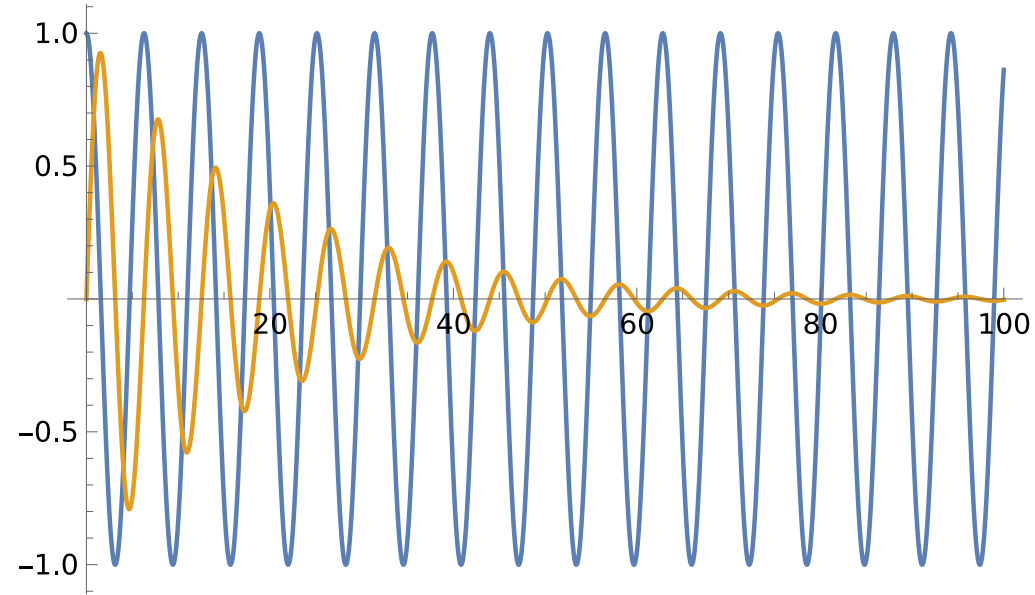
$$= K_1 \cos\left(t \frac{1}{\sqrt{LC}}\right) + K_3 e^{-t \frac{1}{RC}} e^{t((RC)^{-2} - (LC)^{-1})^{1/2}} + K_4 e^{-t \frac{1}{RC}} e^{-t((RC)^{-2} - (LC)^{-1})^{1/2}}$$

also assuming  $K_3 = K_4$ ,  $K_4$  is real, and  $(RC)^{-2} < (LC)^{-1}$

$$= K_1 \cos\left(t \frac{1}{\sqrt{LC}}\right) + K_3 e^{-t \frac{1}{RC}} \cos\left(t \sqrt{(RC)^{-2} - (LC)^{-1}}\right)$$

# Classical RLLCC Node Analysis Example

Plot of possible solutions.

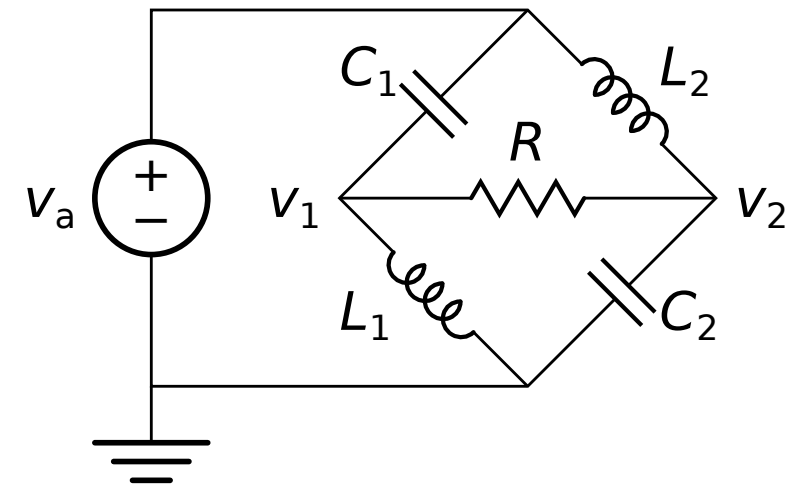


Still need to determine constants (the  $K_1, \dots$ ).

But not for this example ...

... because it won't be necessary with Laplace transforms.

See the Laplace RLLCC section for an analysis with initial conditions.



## Laplace RLC Circuit Node Analysis Example

### Notes

In the time domain (the normal way) ...  
 ... the node voltage function is  $v_1(t)$  ...  
 ... though for brevity it is usually written  $v_1$  (the  $(t)$  is dropped).

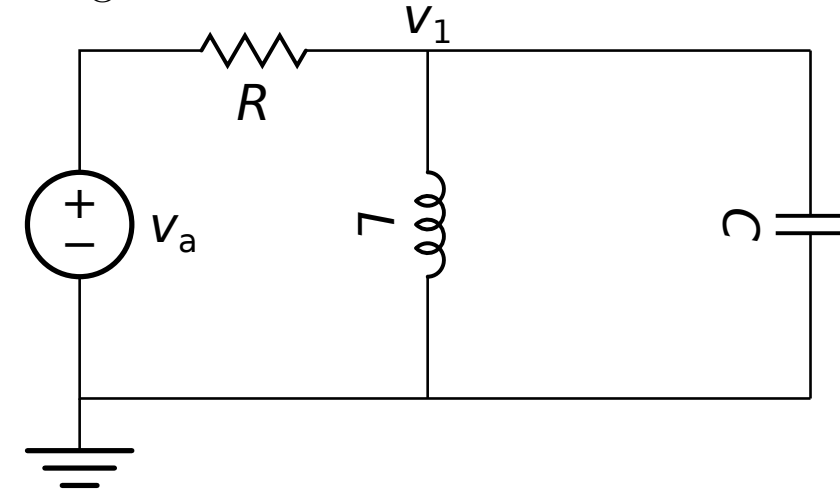
In the  $s$  (complex frequency) domain (the Laplace way) ...  
 ... the node voltage function is  $\mathbf{V}_1(s)$  ...  
 ... though for brevity it is usually written  $\mathbf{V}_1$  (the  $(s)$  is dropped).

The initial capacitor voltage is  $v_C(0^-)$  ...  
 ... and the initial inductors current is  $i_L(0^-)$ .

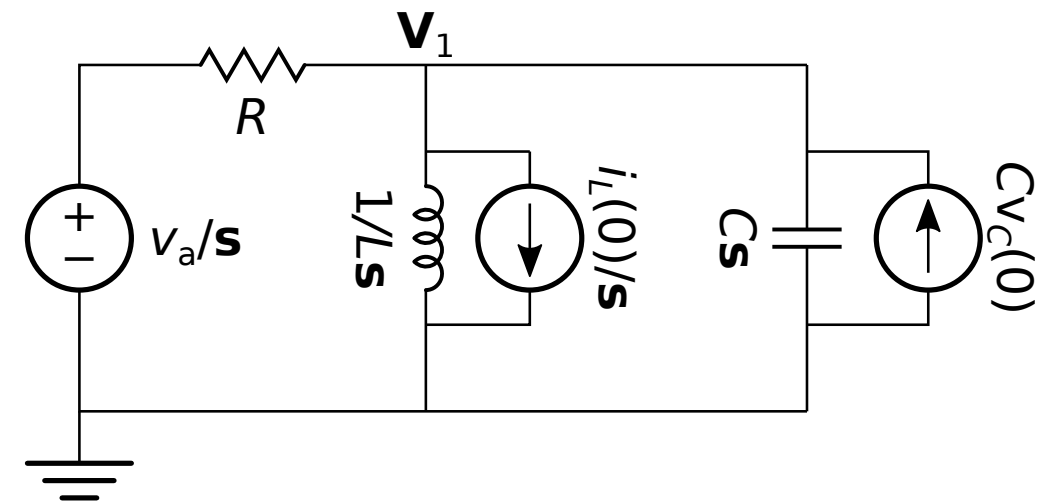
Since capacitor voltage & inductor current can't change instantly...

...  $v_C(0^-) = v_C(0^+)$  and  $i_L(0^-) = i_L(0^+)$  ...  
 ... for brevity they will be shown as  $v_C(0)$  and  $i_L(0)$ .

Original Circuit



Node Analysis Transform-Domain Circuit



## Laplace RLC Circuit Node Analysis Example

Node Equation:

$$\frac{1}{R}(\mathbf{V}_1 - \frac{v_a}{s}) + \frac{1}{Ls}\mathbf{V}_1 + \frac{i_L(0)}{s} + Cs\mathbf{V}_1 - Cv_C(0) = 0$$

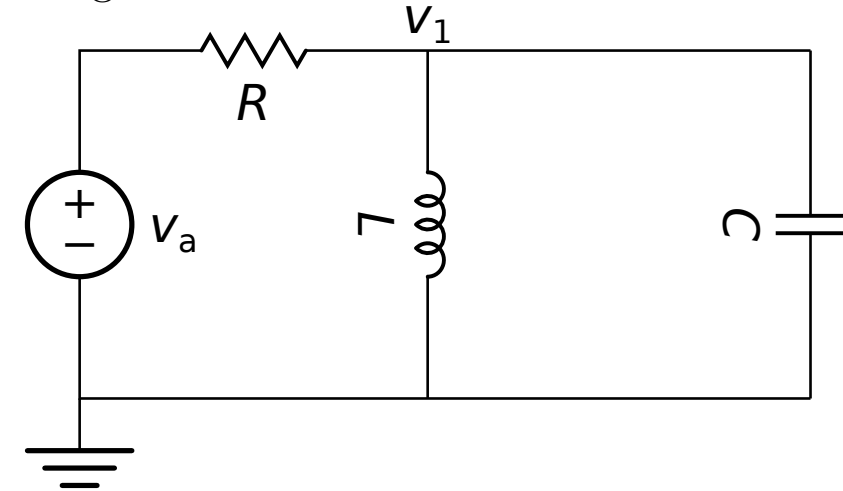
Move knowns to right-hand side, collect  $\mathbf{V}_1$  coefficients on left:

$$\left(Cs + \frac{1}{R} + \frac{1}{Ls}\right)\mathbf{V}_1 = \frac{1}{R}\frac{v_a}{s} - \frac{i_L(0)}{s} + Cv_C(0)$$

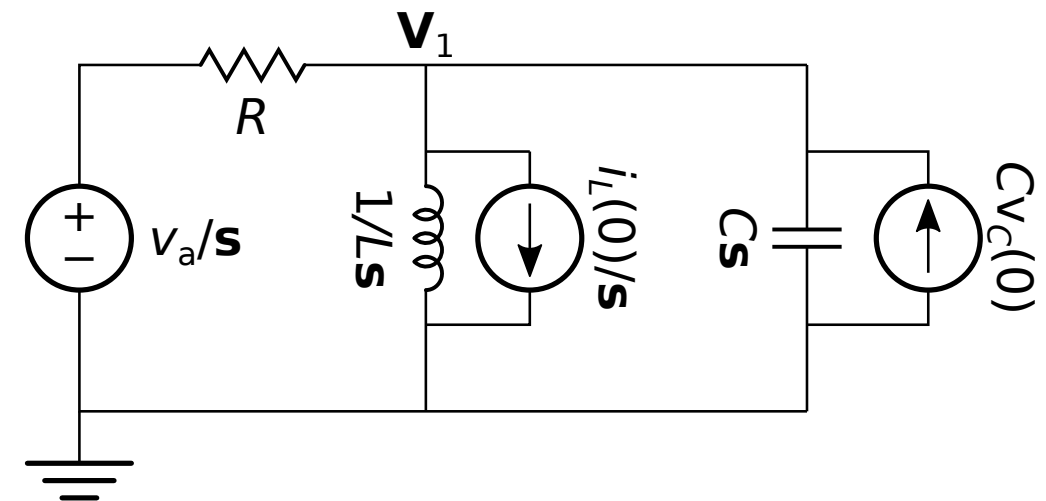
Complete normal form by multiplying by  $\frac{s}{C}$  (to remove  $\frac{1}{s}$  terms and so that coefficient of highest power of  $s$  is 1).

$$\left(s^2 + \frac{1}{RC}s + \frac{1}{LC}\right)\mathbf{V}_1 = \frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0)s$$

Original Circuit



Node Analysis Transform-Domain Circuit





Next, solve for  $\mathbf{V}_1$ :

$$\mathbf{V}_1 = \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0)\mathbf{s}}{\mathbf{s}^2 + \frac{1}{RC}\mathbf{s} + \frac{1}{LC}}$$

then factor the denominator:

$$\mathbf{V}_1 = \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0)\mathbf{s}}{(\mathbf{s} - a_1)(\mathbf{s} - a_2)}$$

where

$$a_1 = -\frac{1}{2RC} + \sqrt{\left(\frac{1}{2RC}\right)^2 - \frac{1}{LC}} \quad a_2 = -\frac{1}{2RC} - \sqrt{\left(\frac{1}{2RC}\right)^2 - \frac{1}{LC}}$$

Next, manipulate  $\mathbf{V}_1$  so that it is a sum of terms of the form  $\frac{K_i}{\mathbf{s} - a_i}$ :

$$\mathbf{V}_1(\mathbf{s}) = \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0)\mathbf{s}}{(\mathbf{s} - a_1)(\mathbf{s} - a_2)} = \begin{cases} \frac{K_1}{\mathbf{s} - a_1} + \frac{K_2}{\mathbf{s} - a_2}, & \text{if } a_1 \neq a_2; \\ \frac{K_1}{(\mathbf{s} - a_1)^2} + \frac{K_2}{\mathbf{s} - a_1}, & \text{if } a_1 = a_2. \end{cases}$$

Time-Domain Solution:

$$v_1(t) = \begin{cases} K_1 e^{a_1 t} + K_2 e^{a_2 t}, & \text{if } a_1 \neq a_2; \\ K_1 t e^{a_1 t} + K_2 e^{a_1 t}, & \text{if } a_1 = a_2. \end{cases}$$

All that remains to do is to find  $K_1$  and  $K_2 \dots$

$\dots$  use partial-fraction expansion to do so.

For the case where  $a_1 \neq a_2$ :

$$\begin{aligned}
 K_1 &= (\mathbf{s} - a_1) \mathbf{V}_1 \Big|_{\mathbf{s}=a_1} \\
 &= (\mathbf{s} - a_1) \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0) \mathbf{s}}{(\mathbf{s} - a_1)(\mathbf{s} - a_2)} \Big|_{\mathbf{s}=a_1} \\
 &= \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0) \mathbf{s}}{(\mathbf{s} - a_2)} \Big|_{\mathbf{s}=a_1} \\
 &= \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0) a_1}{(a_1 - a_2)}
 \end{aligned}$$

Similarly

$$K_2 = \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0) a_2}{(a_2 - a_1)}$$

The s-Domain Solution for the case  $a_1 \neq a_2$ :

Remember that:

$$K_1 = \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0) a_1}{(a_1 - a_2)} \quad K_2 = \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0) a_2}{(a_2 - a_1)}$$

Put the  $K_i$ 's in the expression for  $\mathbf{V}_1$ :

$$\begin{aligned} \mathbf{V}_1 &= \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0) \mathbf{s}}{(\mathbf{s} - a_1)(\mathbf{s} - a_2)} \\ &= \frac{K_1}{\mathbf{s} - a_1} + \frac{K_2}{\mathbf{s} - a_2} \\ &= \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0) a_1}{(a_1 - a_2)(\mathbf{s} - a_1)} + \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0) a_2}{(a_2 - a_1)(\mathbf{s} - a_2)} \end{aligned}$$

For the case where  $a_1 = a_2$ :

$$\mathbf{V}_1 = \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0)\mathbf{s}}{(\mathbf{s} - a_1)^2} = \frac{K_1}{(\mathbf{s} - a_1)^2} + \frac{K_2}{\mathbf{s} - a_1}$$

$$\begin{aligned} K_1 &= (\mathbf{s} - a_1)^2 \mathbf{V}_1 \Big|_{\mathbf{s}=a_1} \\ &= (\mathbf{s} - a_1)^2 \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0)\mathbf{s}}{(\mathbf{s} - a_1)^2} \Big|_{\mathbf{s}=a_1} \\ &= \frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0)\mathbf{s} \Big|_{\mathbf{s}=a_1} \\ &= \frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0)a_1 \end{aligned}$$

## Laplace RLC Circuit Node Analysis Example

$$\begin{aligned} K_2 &= \frac{d}{ds} \left[ (\mathbf{s} - a_1)^2 \mathbf{V}_1 \right] \Big|_{\mathbf{s}=a_1} \\ &= \frac{d}{ds} \left[ (\mathbf{s} - a_1)^2 \frac{\frac{1}{RC} v_a - \frac{i_L(0)}{C} + v_C(0) \mathbf{s}}{(\mathbf{s} - a_1)^2} \right] \Big|_{\mathbf{s}=a_1} \\ &= \frac{d}{ds} \left[ \frac{1}{RC} v_a - \frac{i_L(0)}{C} + v_C(0) \mathbf{s} \right] \Big|_{\mathbf{s}=a_1} \\ &= v_C(0) \Big|_{\mathbf{s}=a_1} \\ &= v_C(0) \end{aligned}$$

The s-Domain Solution for the case  $a_1 = a_2$ :

We just found that:  $K_1 = \frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0) a_1$        $K_2 = v_C(0)$

Put theses  $K_i$ 's in the expression for  $\mathbf{V}_1$ :

$$\begin{aligned} \mathbf{V}_1 &= \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0) \mathbf{s}}{(\mathbf{s} - a_1)^2} \\ &= \frac{K_1}{(\mathbf{s} - a_1)^2} + \frac{K_2}{\mathbf{s} - a_1} \\ &= \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0) a_1}{(\mathbf{s} - a_1)^2} + \frac{v_C(0)}{(\mathbf{s} - a_1)} \end{aligned}$$

## Summary of Laplace RLC Results

s-Domain Voltage:

$$\mathbf{V}_1(\mathbf{s}) = \begin{cases} \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0)a_1}{(a_1 - a_2)(\mathbf{s} - a_1)} + \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0)a_2}{(a_2 - a_1)(\mathbf{s} - a_2)}, & \text{if } a_1 \neq a_2; \\ \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0)a_1}{(\mathbf{s} - a_1)^2} + \frac{v_C(0)}{(\mathbf{s} - a_1)}, & \text{if } a_1 = a_2. \end{cases}$$

Time-Domain Voltage:

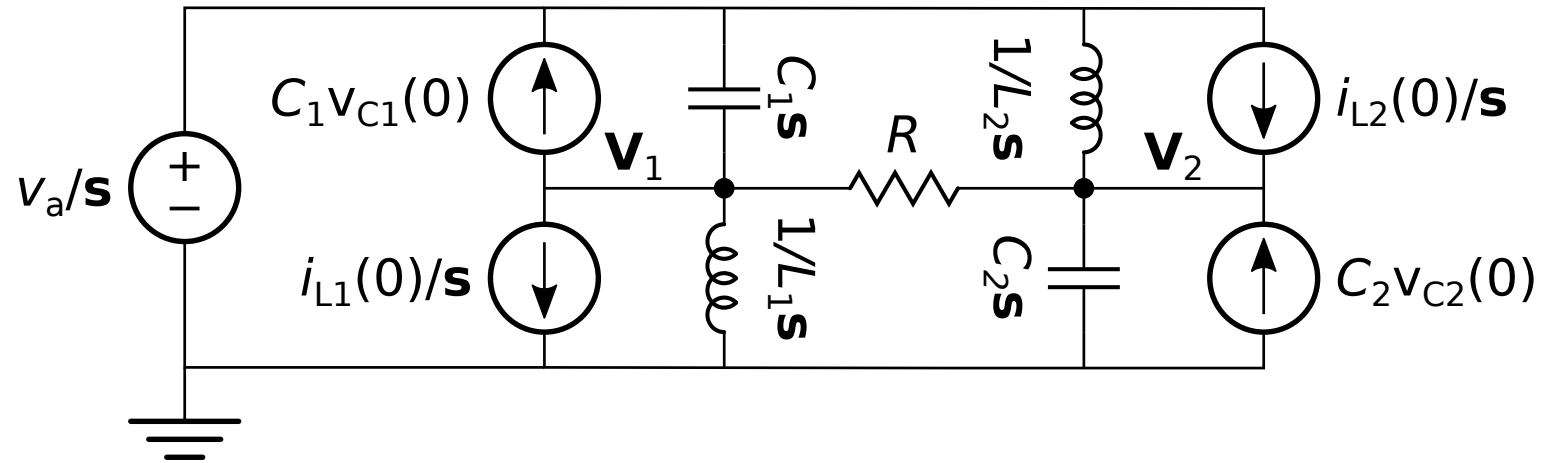
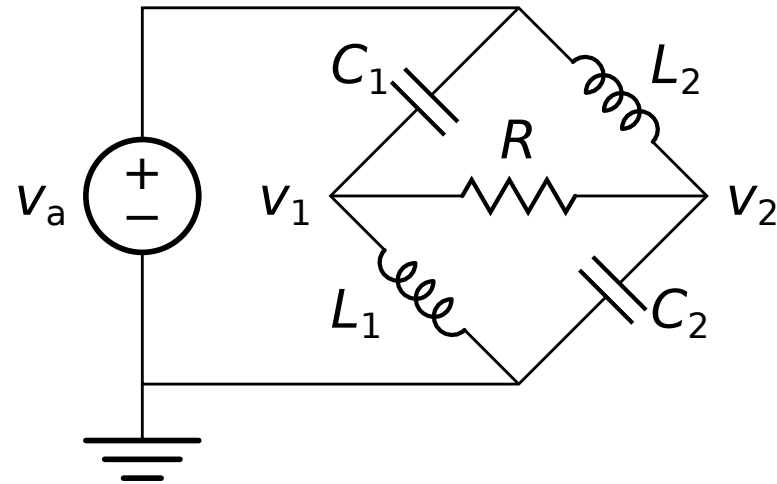
$$v_1(t) = \begin{cases} \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0)a_1}{(a_1 - a_2)} e^{a_1 t} + \frac{\frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0)a_2}{(a_2 - a_1)} e^{a_2 t}, & \text{if } a_1 \neq a_2; \\ \left( \frac{1}{RC}v_a - \frac{i_L(0)}{C} + v_C(0)a_1 \right) t e^{a_1 t} + v_C(0) e^{a_1 t}, & \text{if } a_1 = a_2. \end{cases}$$

where

$$a_1 = -\frac{1}{2RC} + \sqrt{\left(\frac{1}{2RC}\right)^2 - \frac{1}{LC}} \quad a_2 = -\frac{1}{2RC} - \sqrt{\left(\frac{1}{2RC}\right)^2 - \frac{1}{LC}}$$



Laplace RLLCC Circuit Node Analysis Example



Raw Node Equations:

$$C_1 s \left( \mathbf{V}_1 - \frac{v_a}{s} \right) + C_1 v_{C_1}(0) + \frac{1}{R} (\mathbf{V}_1 - \mathbf{V}_2) + \frac{1}{L_1 s} \mathbf{V}_1 + \frac{1}{s} i_{L_1}(0) = 0 \quad \text{Node } \mathbf{V}_1$$

$$\frac{1}{L_2 s} \left( \mathbf{V}_2 - \frac{v_a}{s} \right) - \frac{1}{s} i_{L_2}(0) + \frac{1}{R} (\mathbf{V}_2 - \mathbf{V}_1) + C_2 s \mathbf{V}_2 - C_2 v_{C_2}(0) = 0 \quad \text{Node } \mathbf{V}_2$$

Normal Form Step 1:

Move knowns to right-hand side, collect terms multiplying unknowns ( $\mathbf{V}_1$  and  $\mathbf{V}_2$ ).

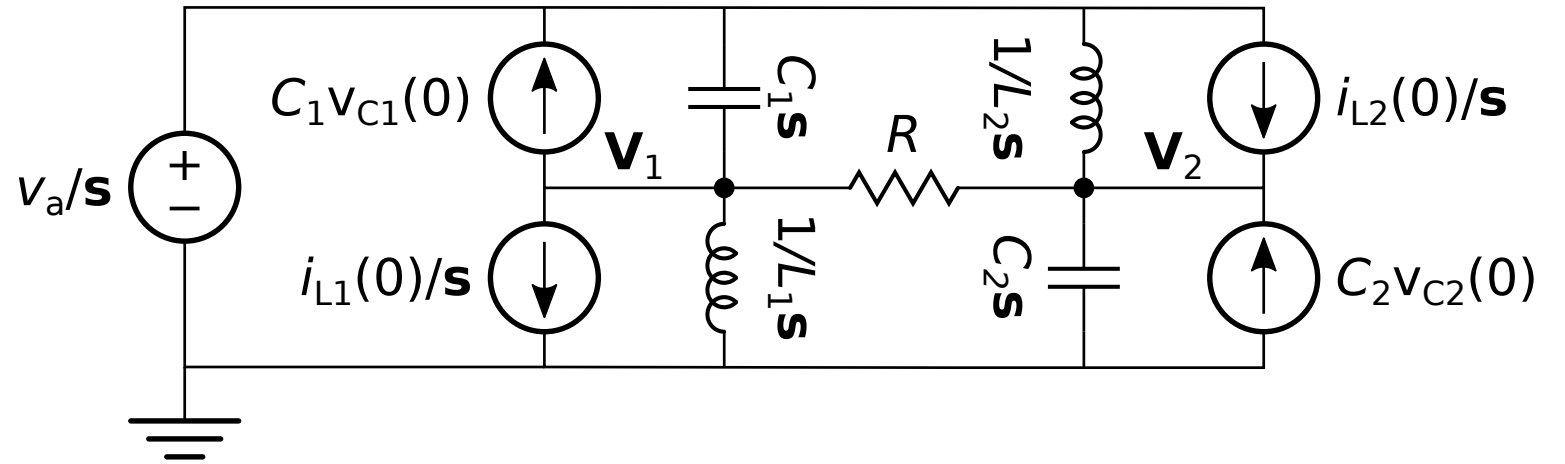
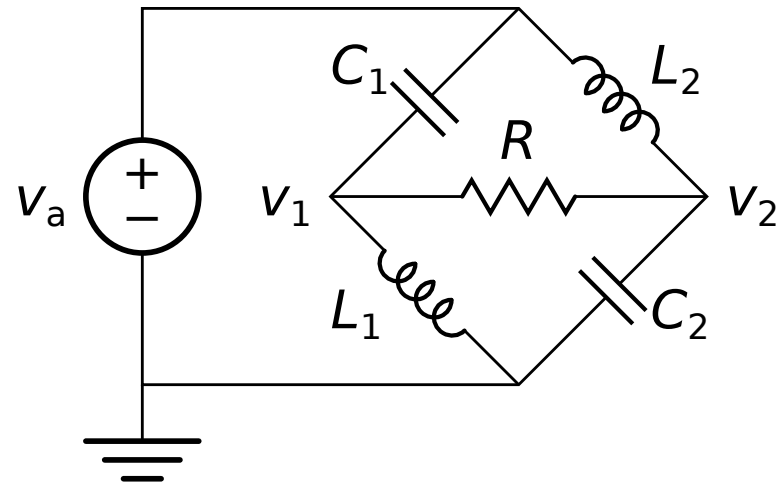
$$\begin{aligned} \left(C_1\mathbf{s} + \frac{1}{R} + \frac{1}{L_1\mathbf{s}}\right)\mathbf{V}_1 + \left(-\frac{1}{R}\right)\mathbf{V}_2 &= C_1v_a - C_1v_{C_1}(0) - \frac{1}{\mathbf{s}}i_{L_1}(0) && \text{Node } \mathbf{V}_1 \\ \left(-\frac{1}{R}\right)\mathbf{V}_1 + \left(C_2\mathbf{s} + \frac{1}{R} + \frac{1}{L_2\mathbf{s}}\right)\mathbf{V}_2 &= \frac{1}{L_2\mathbf{s}^2}v_a + C_2v_{C_2}(0) + \frac{1}{\mathbf{s}}i_{L_2}(0) && \text{Node } \mathbf{V}_2 \end{aligned}$$

Normal Form Step 2:

Multiply by  $\mathbf{s}$  until there are no more  $\frac{1}{\mathbf{s}}$  terms.

$$\begin{aligned} \left(C_1\mathbf{s}^2 + \frac{1}{R}\mathbf{s} + \frac{1}{L_1}\right)\mathbf{V}_1 + \left(-\frac{1}{R}\mathbf{s}\right)\mathbf{V}_2 &= C_1\mathbf{s}v_a - C_1\mathbf{s}v_{C_1}(0) - i_{L_1}(0) && \text{Node } \mathbf{V}_1 \\ \left(-\frac{1}{R}\mathbf{s}^2\right)\mathbf{V}_1 + \left(C_2\mathbf{s}^3 + \frac{1}{R}\mathbf{s}^2 + \frac{1}{L_2}\mathbf{s}\right)\mathbf{V}_2 &= \frac{1}{L_2}v_a + C_2\mathbf{s}^2v_{C_2}(0) + \mathbf{s}i_{L_2}(0) && \text{Node } \mathbf{V}_2 \end{aligned}$$

Laplace RLLCC Circuit Node Analysis Example



Solving simultaneously and cleaning up a bit ...

$$\mathbf{V}_1 = \frac{\mathbf{s}^3(v_a - v_{C_1}(0)) + \mathbf{s}^2 \left( \frac{v_{C_2}(0)}{RC_1} - \frac{v_{C_1}(0)}{RC_2} - \frac{i_{L_1}(0)}{C_1} + \frac{v_a}{RC_2} \right) + \mathbf{s} \left( \frac{i_{L_2}(0)}{RC_1 C_2} - \frac{i_{L_1}(0)}{RC_1 C_2} + \frac{v_a}{C_2 L_2} - \frac{v_{C_1}(0)}{C_2 L_2} \right) + \left( \frac{v_a}{RL_2 C_1 C_2} - \frac{i_{L_1}(0)}{L_2 C_1 C_2} \right)}{\mathbf{s}^4 + \mathbf{s}^3 \left( \frac{1}{RC_2} + \frac{1}{RC_1} \right) + \mathbf{s}^2 \left( \frac{1}{L_1 C_1} + \frac{1}{L_2 C_2} \right) + \mathbf{s} \left( \frac{1}{RL_1 C_1 C_2} + \frac{1}{RL_2 C_1 C_2} \right) + \frac{1}{L_1 L_2 C_1 C_2}}$$

$$\mathbf{V}_2 = \frac{\mathbf{s}^4 v_{C_2}(0) + \mathbf{s}^3 \left( \frac{i_{L_2}(0)}{C_2} - \frac{(v_{C_1}(0) - v_a)}{RC_2} + \frac{v_{C_2}(0)}{RC_1} \right) + \mathbf{s}^2 \left( \frac{v_a}{L_2 C_2} + \frac{v_{C_2}(0)}{C_1 L_1} - \frac{i_{L_1}(0) - i_{L_2}(0)}{RC_1 C_2} \right) + \frac{\mathbf{s}}{C_1 C_2} \left( \frac{i_{L_2}(0)}{L_1} + \frac{v_a}{RL_2} \right) + \frac{v_a}{L_1 L_2 C_1 C_2}}{\mathbf{s}^5 + \mathbf{s}^4 \left( \frac{1}{RC_1} + \frac{1}{RC_2} \right) + \mathbf{s}^3 \left( \frac{1}{L_1 C_1} + \frac{1}{L_2 C_2} \right) + \mathbf{s}^2 \left( \frac{1}{RL_1 C_1 C_2} + \frac{1}{RL_2 C_1 C_2} \right) + \mathbf{s} \frac{1}{L_1 L_2 C_1 C_2}}$$

Consider the special case  $C_1 \rightarrow C_2 \rightarrow C$  and  $L_1 \rightarrow L_2 \rightarrow L$ :

$$\mathbf{V}_1 = \frac{\mathbf{s}^3 C^2 (v_a - v_{C_1(0)}) + \mathbf{s}^2 C \left( \frac{1}{R} (v_a - v_{C_1(0)} + v_{C_2(0)}) - i_{L_1(0)} \right) + \mathbf{s} \left( \frac{C}{L} (v_a - v_{C_1(0)}) - \frac{i_{L_1(0)}}{R} + \frac{i_{L_2(0)}}{R} \right) + \left( \frac{v_a}{LR} - \frac{i_{L_1(0)}}{L} \right)}{\left( \mathbf{s} + \sqrt{-\frac{1}{LC}} \right) \left( \mathbf{s} - \sqrt{-\frac{1}{LC}} \right) \left( \mathbf{s} + \frac{1}{RC} + \sqrt{\frac{1}{(RC)^2} - \frac{1}{LC}} \right) \left( \mathbf{s} + \frac{1}{RC} - \sqrt{\frac{1}{(RC)^2} - \frac{1}{LC}} \right)}$$

$$\mathbf{V}_2 = \frac{-\mathbf{s}^4 RL^2 C^2 v_{C_2(0)} + \mathbf{s}^3 L^2 C (v_{C_1(0)} - v_{C_2(0)} - i_{L_2(0)} R - v_a) + \mathbf{s}^2 (L^2 (i_{L_1(0)} - i_{L_2(0)}) - RLC (v_a + v_{C_2(0)})) - \mathbf{s} L (i_{L_2(0)} R + v_a) - R v_a}{\mathbf{s} \left( \mathbf{s} + \sqrt{-\frac{1}{LC}} \right) \left( \mathbf{s} - \sqrt{-\frac{1}{LC}} \right) \left( \mathbf{s} + \frac{1}{RC} + \sqrt{\frac{1}{(RC)^2} - \frac{1}{LC}} \right) \left( \mathbf{s} + \frac{1}{RC} - \sqrt{\frac{1}{(RC)^2} - \frac{1}{LC}} \right)}$$

The solutions above are the same as those obtained using the classical analysis, except that there are no constants to find!

Remaining steps:

Use partial-fraction expansion to transform to  $\mathbf{V}_1$  to  $\frac{K_1}{\mathbf{s}-a_1} + \frac{K_2}{\mathbf{s}-a_2} + \frac{K_3}{\mathbf{s}-a_3} + \frac{K_4}{\mathbf{s}-a_4}$  or  $\frac{K_1}{\mathbf{s}-a_1} + \frac{K_2}{\mathbf{s}-a_2} + \frac{K_{32}}{(\mathbf{s}-a_3)^2} + \frac{K_{31}}{\mathbf{s}-a_3}$  form ...  
 ... and  $\mathbf{V}_2$  to  $\frac{K_0}{\mathbf{s}} + \frac{K_1}{\mathbf{s}-a_1} + \frac{K_2}{\mathbf{s}-a_2} + \frac{K_3}{\mathbf{s}-a_3} + \frac{K_4}{\mathbf{s}-a_4}$  or  $\frac{K_0}{\mathbf{s}} + \frac{K_1}{\mathbf{s}-a_1} + \frac{K_2}{\mathbf{s}-a_2} + \frac{K_{32}}{(\mathbf{s}-a_3)^2} + \frac{K_{31}}{\mathbf{s}-a_3}$  form.

Convert to time domain (if desired).

These remaining steps are left as an exercise to the reader.

# Laplace RLLCC Circuit Node Analysis Example

