

EE 7715
RNS Hardware

① e

Consider an RNS system defined by the moduli set $S = \{m_1, m_2, \dots, m_L\}$;
 $(m_i, m_j) = 1$ for $i \neq j$. Such a system will consist of:

- A. Weighted (binary) - to - RNS conversion hardware.
- B. RNS processing hardware.
- C. RNS - to - weighted (binary) conversion hardware.

The hardware organization of the entire RNS system is shown by the figure on next page.

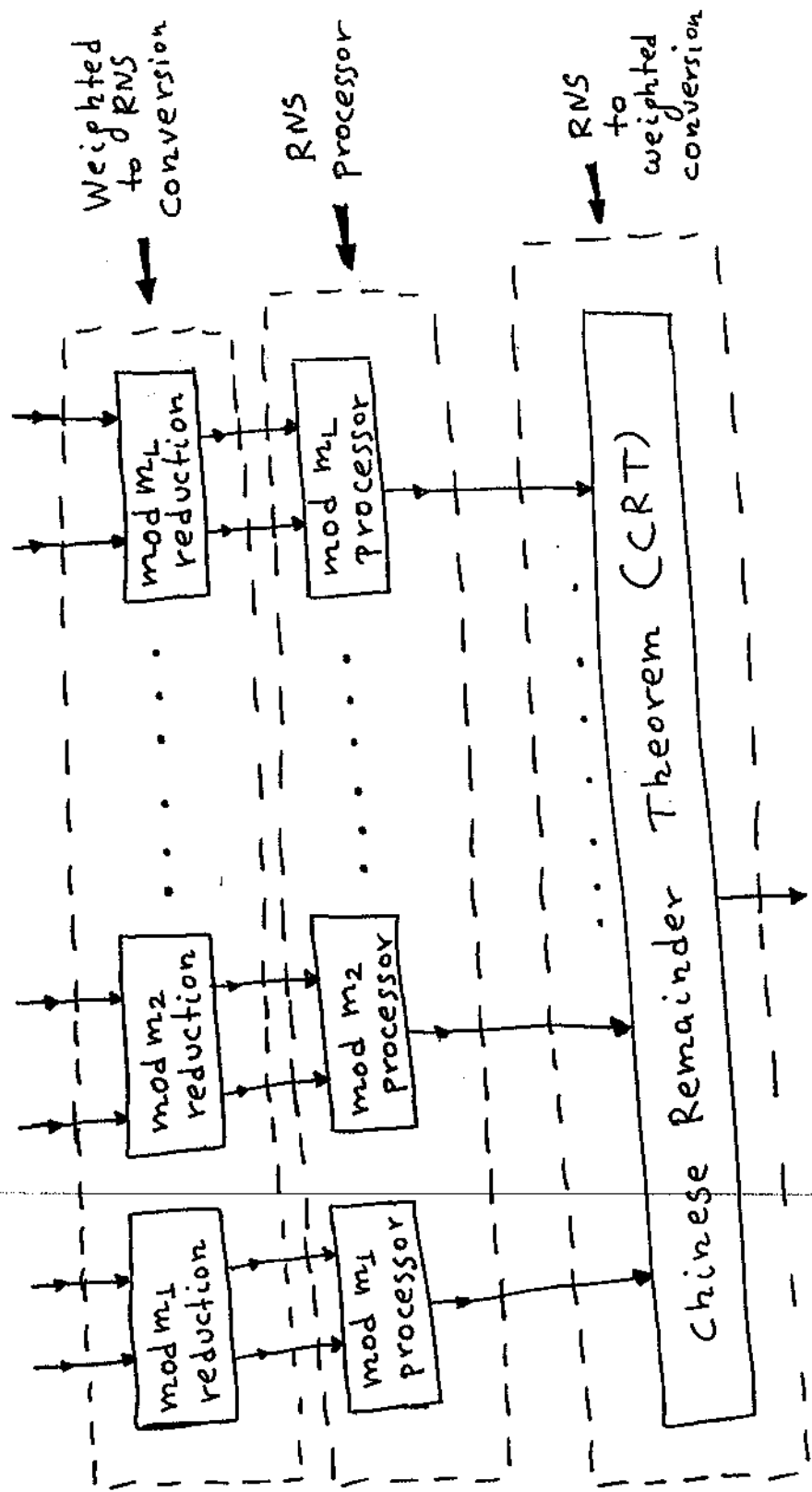


Figure 1: Hardware organization of the entire RNS system using set $S = \{m_1, m_2, \dots, m_L\}$.

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A brief discussion on weighted-to-RNS conversion, RNS processing, and RNS-to-weighted conversion follows.

A. Weighted-to-RNS conversion

The problem here is to evaluate $\langle X \rangle_m$ where X is an n -bit integer while the modulus m is a positive integer.

The evaluation of $\langle X \rangle_m$ can be accomplished by performing the division of X by m and keeping the remainder. This is not a good way, however, since division is a slow operation.

A better technique follows:

Let the n -bit number X be

$$X = (x_{n-1} x_{n-2} \dots x_1 x_0)_2$$

Then

$$X = x_{n-1} \cdot 2^{n-1} + x_{n-2} \cdot 2^{n-2} + \dots + x_1 \cdot 2 + x_0$$

Thus

(4)e

$$\langle X \rangle_m = \langle x_{n-1} \cdot \langle 2^{n-1} \rangle_m + x_{n-2} \cdot \langle 2^{n-2} \rangle_m + \dots + \dots + x_1 \langle 2 \rangle_m + x_0 \rangle_m \quad (1)$$

The terms $\langle 2^i \rangle_m$ can be precomputed and stored; (these terms are preknown constants). A multioperand addition mod m can then complete the computation $\langle X \rangle_m$ of equation (1).

Example: Compute $\langle X \rangle_m$ where $m=19$ and $X=(1111111100)_2$.

$$\text{Here } X = 2^9 + 2^8 + 2^7 + 2^6 + 2^5 + 2^4 + 2^3 + 2^2.$$

Thus,

$$\begin{aligned} \langle X \rangle_m &= \langle \langle 2^9 \rangle_{19} + \langle 2^8 \rangle_{19} + \langle 2^7 \rangle_{19} + \langle 2^6 \rangle_{19} + \\ &\quad + \langle 2^5 \rangle_{19} + \langle 2^4 \rangle_{19} + \langle 2^3 \rangle_{19} + \langle 2^2 \rangle_{19} \rangle_{19} \\ &= \langle -1 + 9 - 5 + 7 + 13 - 3 + 8 + 4 \rangle_{19} = \langle 32 \rangle_{19} = 13. \end{aligned}$$

Double check to see that $\langle X \rangle_m =$

$$= \langle 1020 \rangle_{19} = 13.$$

B. RNS Processing (Arithmetic mod m) (5)e

B-1. Addition modulo m

Let A, B be integers such that $A \in \mathbb{Z}_m$, $B \in \mathbb{Z}_m$ where $\mathbb{Z}_m = \{0, 1, 2, \dots, m-1\}$. The problem here is to compute $\langle A+B \rangle_m$.

Let A and B be n -bit numbers or

$$A = (a_{n-1} a_{n-2} \dots a_1 a_0)_2$$

$$B = (b_{n-1} b_{n-2} \dots b_1 b_0)_2$$

Let S be the summation of A and B
or

$$S = A + B = (s_n s_{n-1} \dots s_1 s_0)_2$$

Obviously, $0 \leq S \leq m+m-2$ (since $0 \leq A \leq m-1$, $0 \leq B \leq m-1$).

If $0 \leq S \leq m-1$ (which means that no overflow occurred due to addition), then

$$\langle A+B \rangle_m = S$$

If $m \leq S \leq m+m-2$ (which means that overflow occurred), then

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$\langle A+B \rangle_m = S-m = S+2^s$ complement of m
(ignore carry out of this addition).

B-2. Additive inverse mod m (negation mod m)

Let A be an integer such that $A \in \mathbb{Z}_m$.

Obviously

$\langle -A \rangle_m = \langle m-A \rangle_m = m+2^s$ compl. of A
(ignore carry out).

B-3. Subtraction mod m

Let A, B be integers such that $A \in \mathbb{Z}_m$,
 $B \in \mathbb{Z}_m$. Then

$$\langle A-B \rangle_m = \langle A + \langle -B \rangle_m \rangle_m.$$

B-4. Multiplication mod m

Let A, B be n -bit integers such that
 $A \in \mathbb{Z}_m$, $B \in \mathbb{Z}_m$. Let P be the full preci-
sion product of A and B or

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$P = A \times B$. Since A, B are n -bit numbers, the product P is a $2n$ -bit number.

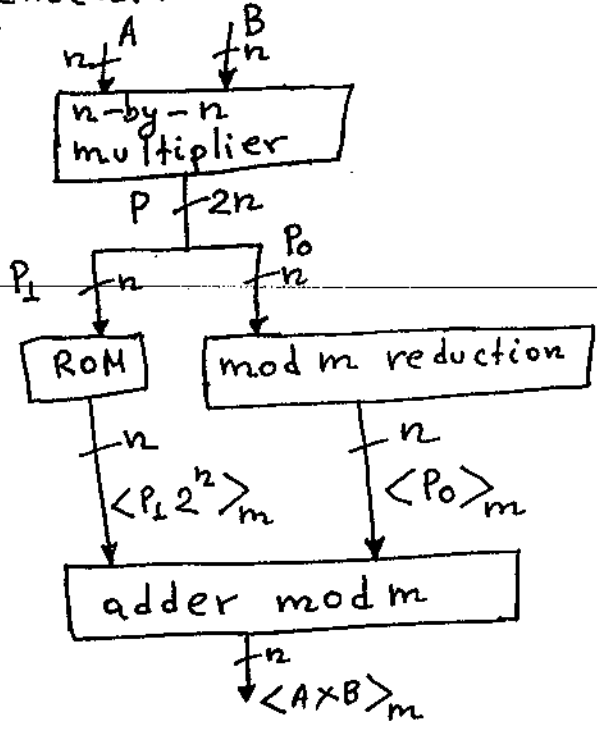
Let P be decomposed into two n -bit blocks as $P = P_1 P_0$. Obviously

$$P = P_1 \cdot 2^n + P_0.$$

Thus

$$\begin{aligned} \langle A \times B \rangle_m &= \langle P \rangle_m = \langle P_1 \cdot 2^n + P_0 \rangle_m = \\ &= \langle \langle P_1 \cdot 2^n \rangle_m + \langle P_0 \rangle_m \rangle_m \quad (2) \end{aligned}$$

The above eq. (2) suggests the following possible implementation.



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Another technique used for mod m multiplication could be the index calculus technique (generators etc), in case where m is prime.

C. RNS-to-weighted conversion using CRT

Let the moduli set be
 $S = \{m_1, m_2, \dots, m_L\}$; $(m_i, m_j) = 1$ for $i \neq j$
 and let the RNS representation of Z be
 $Z \xrightarrow{\text{RNS}} (Z_1, Z_2, \dots, Z_L)$.

The CRT then dictates

$$Z = \left\langle \left\langle Z_1 N_1 \right\rangle_{m_1} \cdot M_1 + \left\langle Z_2 N_2 \right\rangle_{m_2} \cdot M_2 + \dots + \left\langle Z_L N_L \right\rangle_{m_L} \cdot M_L \right\rangle_M$$

where $M = \prod_{i=1}^L m_i$; $M_i = \frac{M}{m_i}$; $N_i = \left\langle M_i^{-1} \right\rangle_{m_i}$

In the above, $M_1, M_2, \dots, M_L, N_1, N_2, \dots, N_L, M$ are preknown constants.

The operations $\left\langle Z_i \cdot N_i \right\rangle_{m_i}$ can be performed using the existing mod m_i multipliers which belong to the RNS processing hardware. Alternatively speaking, $\left\langle Z_i \cdot N_i \right\rangle_{m_i}$ can be performed using special purpose ~~hardware~~ ~~processing~~

multipliers performing
 $\langle Z_i \times \text{constant} \rangle_{m_i}$.

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Post multiplying $\langle Z_i \cdot N_i \rangle_{m_i}$ by M_i can also be implemented by special purpose multipliers performing multiplications by constants.

Finally, an L -operand mod M adder is needed to perform the final CRT addition.

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The 3-moduli RNS based on
set $S = \{2^n - 1, 2^n, 2^n + 1\}$

A residue number system that has been given considerable attention by several ~~more~~ researchers is the 3-moduli RNS based on set

$$S = \{2^n - 1, 2^n, 2^n + 1\}; \quad n = \text{integer.}$$

Such RNS systems imply simple weighted-to-RNS and RNS-to-weighted conversions, simple RNS arithmetic and very balanced decomposition of the dynamic range.

Lemma: Let a be an odd integer. Then the numbers $a, a+1, a+2$ are pairwise relatively prime.

Proof:

Let d be a common divisor of $a+2$ and $a+1$. Then d must divide their difference, or $d \mid 1$ which means that $d=1$. Thus, the only common divisor of $a+2$ and $a+1$ is 1 and $(a+2, a+1) = 1$. The fact that $a+1$

and a are relatively prime can be proven in a similar way. ② f

Regarding the numbers $a+2$ and a , let d' be a common divisor of them. Then d' must divide their difference $a+2-a=2$, or $d'|2$ which means that $d'=1$ or 2 . The number $d'=2$, however, does not divide $a+2$ nor a since both $a+2$ and a are odd numbers. Thus $d'=1$ and $(a+2, a)=1$. The proof is completed.

The above general lemma ~~dictates~~ dictates that the set $S = \{2^n - 1, 2^n, 2^n + 1\}$ consists of pairwise relatively prime integers.

A. Weighted-to-RNS conversion based on set

$$S = \{m_1, m_2, m_3\} = \{2^n - 1, 2^n, 2^n + 1\}$$

An RNS based on set $S = \{m_1, m_2, m_3\} = \{2^n - 1, 2^n, 2^n + 1\}$ achieves a dynamic range of approximately $3n$ bits (see that $M = m_1 \cdot m_2 \cdot m_3 = 2^{3n} - 2^n$).

Let X be an integer such that $X \in \mathbb{Z}_M$ (~~\mathbb{Z}_M~~ $\mathbb{Z}_M = \{0, 1, 2, \dots, M-1\}$).

The residues that need to be computed here ^{(3) f}
 are $X_1 = \langle X \rangle_{2^{n-1}}$, $X_2 = \langle X \rangle_{2^n}$, $X_3 = \langle X \rangle_{2^{n+1}}$.

Let the $3n$ -bit integer X be decomposed
 into three n -bit blocks as $X = X_2 X_1 X_0$.

Obviously, $X = X_2 \cdot 2^{2n} + X_1 \cdot 2^n + X_0$.

Thus,

$$\begin{aligned} X_1 &= \langle X \rangle_{2^{n-1}} = \langle X_2 2^{2n} + X_1 2^n + X_0 \rangle_{2^{n-1}} = \\ &= \langle X_2 (2^n)^2 + X_1 2^n + X_0 \rangle_{2^{n-1}} = \langle X_2 + X_1 + X_0 \rangle_{2^{n-1}} \end{aligned}$$

or

$$\boxed{X_1 = \langle X \rangle_{2^{n-1}} = \langle X_2 + X_1 + X_0 \rangle_{2^{n-1}}} \quad (1)$$

$$X_2 = \langle X \rangle_{2^n} = \langle X_2 (2^n)^2 + X_1 2^n + X_0 \rangle_{2^n} = \langle X_0 \rangle_{2^n} = X_0$$

or

$$\boxed{X_2 = \langle X \rangle_{2^n} = X_0} \quad (2)$$

$$\begin{aligned} X_3 &= \langle X \rangle_{2^{n+1}} = \langle X_2 (2^n)^2 + X_1 2^n + X_0 \rangle_{2^{n+1}} = \\ &= \langle X_2 - X_1 + X_0 \rangle_{2^{n+1}} \end{aligned}$$

or

$$\boxed{X_3 = \langle X \rangle_{2^{n+1}} = \langle X_2 - X_1 + X_0 \rangle_{2^{n+1}}} \quad (3)$$

The facts $\langle 2^n \rangle_{2^n-1} = 1$, $\langle 2^n \rangle_{2^n} = 0$ and (4) $\langle 2^n \rangle_{2^n+1} = \langle -1 \rangle_{2^n+1}$ have been taken into account in deriving equations (1), (2) and (3).

Equations (1), (2), (3) demonstrate that the weighted -to- RNS conversion is simple when using an RNS based on set $S = \{2^n-1, 2^n, 2^n+1\}$.

B. RNS Processing (Arithmetic mod (2^n-1) , mod 2^n , mod (2^n+1))

B-1. Arithmetic mod (2^n-1)

B.1.1 Addition mod (2^n-1)

Let A, B be integers such that $A \in \mathbb{Z}_{2^n-1}$, $B \in \mathbb{Z}_{2^n-1}$, where $\mathbb{Z}_{2^n-1} = \{0, 1, 2, \dots, 2^n-2\}$.

Then $\langle A+B \rangle_{2^n-1} =$ result of 1's complement addition of A and B (add A and B and end around carry). The reason for the end around carry is that the weight factor of the carry out is 2^n while $\langle 2^n \rangle_{2^n-1} = 1 = 2^0$.

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B.1.2 Additive inverse mod $(2^n - 1)$; (negation mod $(2^n - 1)$)

Let A be such that $A \in \mathbb{Z}_{2^n - 1}$. A is then an n -bit number or $A = (a_{n-1} a_{n-2} \dots a_1 a_0)_2$. The additive inverse of A is

$$\begin{aligned} \langle -A \rangle_{2^n - 1} &= \langle 2^n - 1 - A \rangle_{2^n - 1} = \langle (\overline{a_{n-1}} \overline{a_{n-2}} \dots \overline{a_1} \overline{a_0}) \rangle_{2^n - 1} \\ &= (\overline{a_{n-1}} \overline{a_{n-2}} \dots \overline{a_1} \overline{a_0}) = 1\text{'s complement of } A. \end{aligned}$$

Thus

$$\langle -A \rangle_{2^n - 1} = 1\text{'s complement of } A.$$

B.1.3 Subtraction mod $(2^n - 1)$

Obviously,

$$\langle A - B \rangle_{2^n - 1} = A + 1\text{'s complement of } B \text{ (end around carry)}.$$

B.1.4 Scaling by power of two (scaling by 2^k)

Let A be such that $A \in \mathbb{Z}_{2^n - 1}$. The problem here is to compute $\langle 2^k \cdot A \rangle_{2^n - 1}$.

Let A be decomposed into two blocks A_1 and A_0 where A_1 is a k -bit block while A_0 is a $(n-k)$ -bit block.

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Then,

$$A = \underbrace{A_1}_{\substack{k\text{-} \\ \text{bit}}} \underbrace{A_0}_{\substack{n-k \\ \text{bit}}} = A_1 \cdot 2^{n-k} + A_0.$$

$$\text{Thus, } \langle 2^k \cdot A \rangle_{2^n-1} = \langle 2^k \cdot (A_1 \cdot 2^{n-k} + A_0) \rangle_{2^n-1}$$

$$= \langle A_1 \cdot 2^n + A_0 \cdot 2^k \rangle_{2^n-1} = \langle A_0 \cdot 2^k + A_1 \rangle_{2^n-1} =$$

$$= \langle \underbrace{(A_0 \text{ } \underbrace{000 \dots 000}_{k \text{ zeros}})}_{\substack{n-k \\ \text{bit}}} + \underbrace{A_1}_{\substack{k \\ \text{bit}}} \rangle_{2^n-1} = \langle A_0 A_1 \rangle_{2^n-1} = A_0 A_1 =$$

= result of rotating A left by k bits.

So

$$\langle 2^k \cdot A \rangle_{2^n-1} = \text{result of left-rotating } A \text{ by } k \text{ bits}$$

Example: Consider $n=8$ and $A = (a_7 a_6 a_5 a_4 a_3 a_2 a_1 a_0)_2$.

$$\text{Then } \langle 2^3 \cdot A \rangle_{2^8-1} = (a_4 a_3 a_2 a_1 a_0 a_7 a_6 a_5)_2.$$

B.1.5 Multiplication mod (2^n-1)

Let A, B be n -bit integers such that $A \in \mathbb{Z}_{2^n-1}, B \in \mathbb{Z}_{2^n-1}$. Here, $\langle A \times B \rangle_{2^n-1}$ needs to be

computed. Let P be the full precision ^⑦ product of A and B or $P = A \times B$. Then P is a $2n$ -bit number which can be decomposed into two n -bit blocks as $P = P_1 P_0$. Obviously $P = P_1 \cdot 2^n + P_0$ and thus

$$\begin{aligned} \langle A \times B \rangle_{2^n - 1} &= \langle P \rangle_{2^n - 1} = \langle P_1 \cdot 2^n + P_0 \rangle_{2^n - 1} \\ &= \langle P_1 + P_0 \rangle_{2^n - 1}. \end{aligned}$$

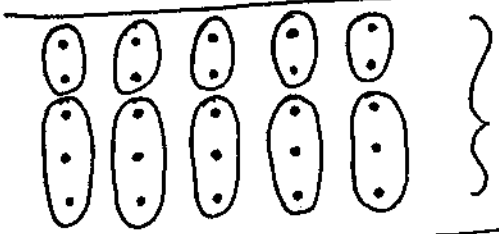
The above equation suggests that $\langle A \times B \rangle_{2^n - 1}$ can be computed by an n -by- n full precision multiplier computing $A \times B = P = P_1 P_0$, followed by a mod $(2^n - 1)$ adder performing $\langle P_1 + P_0 \rangle_{2^n - 1}$.

The above technique is not the most hardware efficient technique. A much better approach is a direct design of a mod $(2^n - 1)$ multiplier based on counters. This approach is shown by an example on next page.

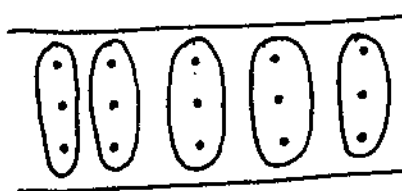
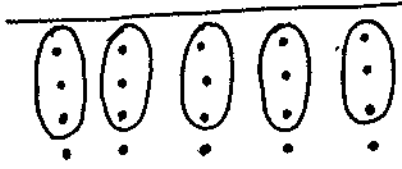
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Let A, B be 5-bit integers such that $A \in \mathbb{Z}_{2^5-1}$, $B \in \mathbb{Z}_{2^5-1}$ and suppose that $\langle A \times B \rangle_{2^5-1}$ needs to be computed. Below you can see a counter-based implementation of $\langle A \times B \rangle_{2^5-1}$. The counters used are (3,2) and (2,2) counters.

$$A = a_4 a_3 a_2 a_1 a_0$$
$$B = b_4 b_3 b_2 b_1 b_0$$



matrix of summands of $\langle A \times B \rangle_{2^5-1}$



mod (2^5-1) CPA (end around carry).



$\langle A \times B \rangle_{2^5-1}$

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B.2 Arithmetic mod 2^n

B.2.1 Addition mod 2^n

Let A, B be such that $A \in \mathbb{Z}_{2^n}$, $B \in \mathbb{Z}_{2^n}$.

Then $\langle A+B \rangle_{2^n}$ = result of 2^n 's complement addition of A and B ; (add A and B and ignore carry out).

The reason for ignoring the carry out is that its weight factor is 2^n while $\langle 2^n \rangle_{2^n} = 0$.

B.2.2 Additive inverse mod 2^n

Let $A \in \mathbb{Z}_{2^n}$. Then

$$\langle -A \rangle_{2^n} = \langle 2^n - A \rangle_{2^n} = 2^n \text{'s complement of } A.$$

B.2.3 Subtraction mod 2^n

$$\langle A-B \rangle_{2^n} = A + 2^n \text{'s complement of } B; \text{ (ignore carry out).}$$

B.2.4 Scaling by power of two (scaling by 2^k)

Let A be such that $A \in \mathbb{Z}_{2^n}$. Decompose A into two blocks A_1 and A_0 where A_1 is a k -bit block while A_0 is a $(n-k)$ -bit block. Then

$$A = A_1 A_0 = A_1 \cdot 2^{n-k} + A_0.$$

$$\begin{aligned}
 \text{Thus, } \langle 2^k \cdot A \rangle_{2^n} &= \langle 2^k \cdot (A_1 \cdot 2^{n-k} + A_0) \rangle_{2^n} = \textcircled{10} f \\
 &= \langle A_1 \cdot 2^n + A_0 \cdot 2^k \rangle_{2^n} = \langle A_0 \cdot 2^k \rangle_{2^n} = \\
 &= \langle (\underbrace{A_0}_{\substack{\uparrow \\ n-k \\ \text{bit}}} \underbrace{000 \dots 000}_{k \text{ zeros}}) \rangle_{2^n} = A_0 000 \dots 000
 \end{aligned}$$

So

$\langle 2^k \cdot A \rangle_{2^n}$ = result of left-shifting A by k bits
zero-filling at the right

Example: Consider $n=8$ and $A = (a_7 a_6 a_5 a_4 a_3 a_2 a_1 a_0)_2$.

$$\text{Then } \langle 2^3 \cdot A \rangle_{2^8} = (a_4 a_3 a_2 a_1 a_0 000)_2.$$

B.2.5 Multiplication mod 2^n

Let $A \in \mathbb{Z}_{2^n}$, $B \in \mathbb{Z}_{2^n}$ and suppose that $\langle A \times B \rangle_{2^n}$ needs to be computed. Let P be $P = A \times B$; (P is the full precision product of A and B). Then $P = \underbrace{P_1}_{\substack{\uparrow \\ n \\ \text{bit}}} \underbrace{P_0}_{\substack{\uparrow \\ n \\ \text{bit}}} = P_1 \cdot 2^n + P_0$.

$$\text{Thus, } \langle A \times B \rangle_{2^n} = \langle P \rangle_{2^n} = \langle P_1 \cdot 2^n + P_0 \rangle_{2^n} = P_0$$

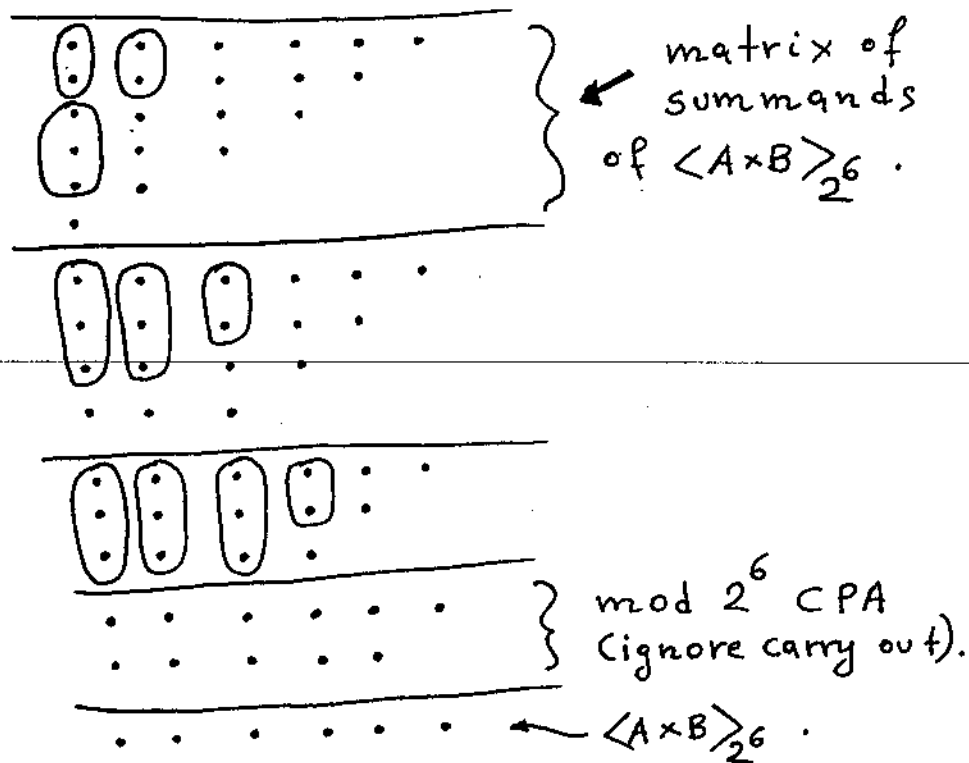
The above approach which is based on computing the full precision product of A and B

and keeping its right most n -bit part P_0 ^{(11) f}
 is not the most hardware efficient approach.
 A much better direct design of a mod 2^n
 multiplier based on counters can be used.
 This is demonstrated by the following example.

Let A, B be 6-bit integers such that
 $A \in \mathbb{Z}_{2^6}$, $B \in \mathbb{Z}_{2^6}$ and suppose that $\langle A \times B \rangle_{2^6}$
 needs to be computed. A counter-based imple-
 mentation of $\langle A \times B \rangle_{2^6}$ is shown below. The
 counters used are $(3, 2)$ and $(2, 2)$ counters.

$$A = a_5 a_4 a_3 a_2 a_1 a_0$$

$$B = b_5 b_4 b_3 b_2 b_1 b_0$$



(12) f

B.3 Arithmetic mod (2^n+1) .

B.3.1 Addition mod (2^n+1)

Let A, B be integers such that $A \in \mathbb{Z}_{2^n+1}$, $B \in \mathbb{Z}_{2^n+1}$ where $\mathbb{Z}_{2^n+1} = \{0, 1, 2, \dots, 2^n\}$. Then A and B need $n+1$ bits for their representation or $A = (a_n a_{n-1} \dots a_1 a_0)_2$; $B = (b_n b_{n-1} \dots b_1 b_0)_2$.

Case (i): $a_n = b_n = 1$: In this case, $A = B = (1000 \dots 00)_2 = 2^n$. Thus $\langle A+B \rangle_{2^n+1} = \langle 2^n + 2^n \rangle_{2^n+1} = \langle (-1) + (-1) \rangle_{2^n+1} = \langle -2 \rangle_{2^n+1} = 2^n + 1 - 2 = 2^n - 1 = (0 \underbrace{111 \dots 111}_n)_2$. This case can be detected and

treated separately.

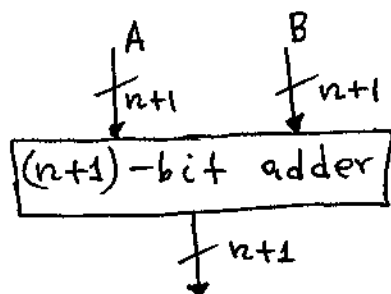
Case (ii): At least one of a_n, b_n is zero:

In this case, at least one of the two numbers A or B is smaller than 2^n .

Let S be the full precision summation of A and B or $S = A+B$. Here, S is an $(n+1)$ -bit number since $S_{\max} = (A+B)_{\max} = 2^n + (2^n - 1) = 2^{n+1} - 1$; (recall that not both numbers A, B are equal to 2^n).

(13) f

The computation $S = A + B$ can be performed by a regular $(n+1)$ -bit binary adder



$$S = (s_n s_{n-1} \dots s_1 s_0)_2$$

As explained $0 \leq S \leq 2^{n+1} - 1$.

- If $0 \leq S \leq 2^n$ (which means that overflow did not occur or $S \in \mathbb{Z}_{2^{n+1}}^n$), then

$$\langle A+B \rangle_{2^{n+1}} = S$$

- If $S > 2^n$ then an overflow occurred ($S \notin \mathbb{Z}_{2^{n+1}}^n$) and in this case

$$\langle A+B \rangle_{2^{n+1}} = \langle S \rangle_{2^{n+1}}$$

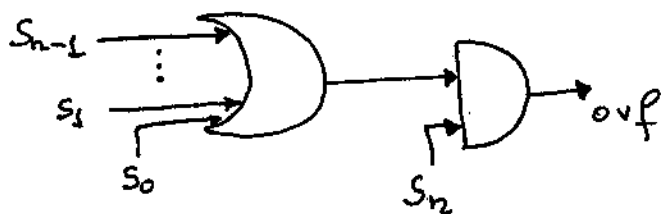
The overflow is easily detected. Overflow occurs if

$$S = (s_n s_{n-1} \dots s_1 s_0)_2 > 2^n.$$

(14) f

This means:

$S_n = 1$ and not all $S_{n-1}, S_{n-2}, \dots, S_1, S_0$ are zeros. As shown below the hardware for overflow detection consists of one OR and one AND gate



In this case of an overflow

$$2^{n+1} \leq S \leq 2^{n+1} - 1$$

or

$$2^{n+1} \leq S \leq 2^n + 2^n - 1$$

Then

$$\langle A+B \rangle_{2^{n+1}} = \langle S \rangle_{2^{n+1}} = S - (2^{n+1}) =$$

$$= (S_n S_{n-1} \dots S_1 S_0)_2 - (2^{n+1}) = 2^n + (S_{n-1} \dots S_1 S_0)_2 - 2^n - 1$$

$$= (S_{n-1} \dots S_1 S_0)_2 - 1 = (S_{n-1} \dots S_1 S_0)_2 + 2^n \text{ compl. of } 1 =$$

$$= (S_{n-1} S_{n-2} \dots S_1 S_0)_2 + (111 \dots 111)_2 \quad (\text{ignore carry out of this addition}).$$

Thus, in this case of $S > 2^n$, the

desired $\langle A+B \rangle_{2^{n+1}}$ can be obtained (15) f
 by adding $(s_{n-1}s_{n-2}\dots s_1s_0)_2$ with a vector
 consisting of n ones and keeping the
 right most n bits of the result.

B.3.2 Additive inverse mod (2^{n+1})

Let A be an integer such that $A \in \mathbb{Z}_{2^{n+1}}$;
 $(\mathbb{Z}_{2^{n+1}} = \{0, 1, 2, \dots, 2^n\})$.

Then $A = (a_n a_{n-1} \dots a_1 a_0)_2$.

Obviously

$$\langle -A \rangle_{2^{n+1}} = \langle 2^{n+1} - A \rangle_{2^{n+1}}$$

Case (i): $A=0$

In this case $\langle -A \rangle_{2^{n+1}} = 0$. This case
 can be detected and treated separately.

Case (ii): $A=2^n$

Here, $\langle -A \rangle_{2^{n+1}} = \langle 2^{n+1} - 2^n \rangle_{2^{n+1}} = 1$. This
 case can be detected and treated
 separately.

(16) f

Case (iii): $0 < A < 2^n$

In this case A can be represented by n bits or $A = (a_{n-1} a_{n-2} \dots a_1 a_0)_2$.

$$\begin{aligned} \text{Then } \langle -A \rangle_{2^{n+1}} &= \langle 2^n + 1 - A \rangle_{2^{n+1}} = \\ &= \langle 2^n + 1 - (a_{n-1} \dots a_1 a_0)_2 \rangle_{2^{n+1}} = \\ &= \langle 2^n - 1 - (a_{n-1} a_{n-2} \dots a_1 a_0)_2 + 2 \rangle_{2^{n+1}} = \\ &= \langle (\overline{a_{n-1}} \overline{a_{n-2}} \dots \overline{a_1} \overline{a_0})_2 + 2 \rangle_{2^{n+1}}. \end{aligned}$$

The fact that $A \neq 0$ implies that

$$(\overline{a_{n-1}} \overline{a_{n-2}} \dots \overline{a_1} \overline{a_0})_2 \leq 2^n - 2 \quad \text{or}$$

$$(\overline{a_{n-1}} \overline{a_{n-2}} \dots \overline{a_1} \overline{a_0})_2 + 2 \leq 2^n. \quad \text{Thus,}$$

$$\begin{aligned} \langle (\overline{a_{n-1}} \overline{a_{n-2}} \dots \overline{a_1} \overline{a_0})_2 + 2 \rangle_{2^{n+1}} &= \\ &= (\overline{a_{n-1}} \overline{a_{n-2}} \dots \overline{a_1} \overline{a_0})_2 + 2. \end{aligned}$$

In conclusion, in this case where $0 < A < 2^n$,

$$\langle -A \rangle_{2^{n+1}} = (\overline{a_{n-1}} \overline{a_{n-2}} \dots \overline{a_1} \overline{a_0})_2 + 2$$

which means that $\langle -A \rangle_{2^{n+1}}$ is the 1's compl. of A incremented by 2.

B.3.3 Subtraction mod (2^n+1)

Let A, B be such that $A \in \mathbb{Z}_{2^n+1}, B \in \mathbb{Z}_{2^n+1}$.

Obviously

$$\langle A - B \rangle_{2^n+1} = \langle A + \langle -B \rangle_{2^n+1} \rangle_{2^n+1}$$

B.3.4 Scaling by power of two (scaling by 2^k)

Let A be such that $A \in \mathbb{Z}_{2^n+1}$. Then

$A = (a_n a_{n-1} \dots a_1 a_0)_2$. The problem here is to compute $\langle 2^k \cdot A \rangle_{2^n+1}$.

Let A be decomposed into two blocks A_1 and A_0 , where A_1 is a $(k+1)$ -bit block while A_0 is a $(n-k)$ -bit block.

Then

$$A = \underbrace{A_1}_{\substack{k+1 \\ \text{bit}}} \underbrace{A_0}_{\substack{n-k \\ \text{bit}}} = A_1 \cdot 2^{n-k} + A_0$$

$$\begin{aligned} \text{Thus, } \langle 2^k \cdot A \rangle_{2^n+1} &= \langle 2^k \cdot (A_1 \cdot 2^{n-k} + A_0) \rangle_{2^n+1} \\ &= \langle A_1 \cdot 2^n + A_0 \cdot 2^k \rangle_{2^n+1} = \langle -A_1 + A_0 \cdot 2^k \rangle_{2^n+1} \end{aligned}$$

(18) f

B.3.5 Multiplication mod (2^n+1)

Let A and B be integers such that $A, B \in \mathbb{Z}_{2^n+1}$. Here $\langle A \times B \rangle_{2^n+1}$ needs to be computed. Let the binary representations of A and B be

$$A = (a_n a_{n-1} \dots a_1 a_0)_2 ; B = (b_n b_{n-1} \dots b_1 b_0)_2.$$

Case (i): $a_n = b_n = 1$:

In this case $A = B = (100 \dots 0)_2 = 2^n = \langle -1 \rangle_{2^n+1}$.

$$\text{Thus } \langle A \times B \rangle_{2^n+1} = \langle (-1) \times (-1) \rangle_{2^n+1} = 1.$$

This case can be detected and treated separately.

Case (ii): $a_n = 1, b_n = 0$:

Here $A = 2^n = \langle -1 \rangle_{2^n+1}$ and $\langle A \times B \rangle_{2^n+1} =$

$= \langle -B \rangle_{2^n+1}$. In this case the necessary

hardware is just a negator mod (2^n+1) .

Case (iii): $a_n = 0, b_n = 1$:

Here $B = 2^n = \langle -1 \rangle_{2^n+1}$ and $\langle A \times B \rangle_{2^n+1} =$

$= \langle -A \rangle_{2^n+1}$.

Case (iv): $a_n = b_n = 0$

(19) f

In this case the integers A and B are n -bit numbers with binary representations $A = (a_{n-1} a_{n-2} \dots a_1 a_0)_2$; $B = (b_{n-1} \dots b_1 b_0)_2$.

Let P be the full precision product of A and B or $P = A \times B$. The product P is a $2n$ -bit number and can be decomposed into two n -bit blocks as $P = P_1 P_0$. Obviously $P = P_1 \cdot 2^n + P_0$ and thus

$$\begin{aligned} \langle A \times B \rangle_{2^{n+1}} &= \langle P \rangle_{2^{n+1}} = \langle P_1 \cdot 2^n + P_0 \rangle_{2^{n+1}} \\ &= \langle -P_1 + P_0 \rangle_{2^{n+1}}. \end{aligned}$$

The above equation suggests that a possible implementation of $\langle A \times B \rangle_{2^{n+1}}$ could rely on an n -by- n full-precision multiplier computing $A \times B = P = P_1 P_0$, followed by a $\text{mod}(2^{n+1})$ subtractor performing $\langle -P_1 + P_0 \rangle_{2^{n+1}}$.

Note:

A more efficient way for performing arithmetic mod $(2^n + 1)$ is by using diminished-1 representations of numbers. Few references on this subject follow:

References

- [1] L.M. Leibowitz, "A simplified binary arithmetic for the fermat number transform", IEEE Trans. Acoustics, Speech, and Signal Processing, vol. ASSP-24, no. 5, pp. 356-359, Oct. 1976.
- [2] Z. Wang, G.A. Jullien and W.C. Miller, "An algorithm for multiplication modulo $(2^n + 1)$ ", in Proceedings of 29th Asilomar Conference on Signals, Systems, and Computers, (Pacific Grove, CA, Oct. 1995), pp. 956-960.

A. The Chinese Remainder Theorem (CRT) for the attractive 3-moduli RNS.

Consider the RNS based on the set

$$S = \{m_1, m_2, m_3\} = \{2^n - 1, 2^n, 2^n + 1\}; \quad n \text{ integ.}$$

Let the integer X be represented in the RNS as $X \xrightarrow{\text{RNS}} (X_1, X_2, X_3)$

Then the CRT reconstructs X from its residues as

$$X = \left\langle \left\langle X_1 N_1 \right\rangle_{m_1} M_1 + \left\langle X_2 N_2 \right\rangle_{m_2} M_2 + \left\langle X_3 N_3 \right\rangle_{m_3} M_3 \right\rangle_M$$

In the above $M = m_1 \times m_2 \times m_3$;

$$M_1 = \frac{M}{m_1} = m_2 m_3 = 2^n (2^n + 1);$$

$$N_1 = \left\langle M_1^{-1} \right\rangle_{m_1} = \left\langle \left[2^n (2^n + 1) \right]^{-1} \right\rangle_{2^n - 1} = \left\langle (1 \times 2)^{-1} \right\rangle_{2^n - 1}$$

$$= \left\langle 2^{-1} \right\rangle_{2^n - 1} = \frac{2^n - 1 + 1}{2} = 2^{n-1}$$

• In the above we used the fact that if m is an odd integer then $\left\langle 2^{-1} \right\rangle_m = \frac{m+1}{2}$; (double check to see that $\left\langle 2 \times \frac{m+1}{2} \right\rangle_m = \left\langle m+1 \right\rangle_m = 1$).

$$M_2 = \frac{M}{m_2} = m_1 m_3 = (2^n - 1)(2^n + 1) \quad (2)g$$

$$N_2 = \langle M_2^{-1} \rangle_{m_2} = \langle [(2^n - 1)(2^n + 1)]^{-1} \rangle_{2^n} = \langle [(-1)(1)]^{-1} \rangle_{2^n}$$

$$= \langle (-1)^{-1} \rangle_{2^n} = \langle -1 \rangle_{2^n}$$

$$M_3 = \frac{M}{m_3} = m_1 \times m_2 = (2^n - 1)2^n$$

$$N_3 = \langle M_3^{-1} \rangle_{m_3} = \langle [(2^n - 1)2^n]^{-1} \rangle_{2^{n+1}} = \langle [(-1)(-1)]^{-1} \rangle_{2^{n+1}}$$

$$= \langle 2^{-1} \rangle_{2^{n+1}} = \frac{2^{n+1} + 1}{2} = 2^{n-1} + 1$$

Then the CRT gives

$$X = \left\langle \left\langle X_1 2^{n-1} \right\rangle_{2^{n-1}} \times 2^n (2^n + 1) + \langle -X_2 \rangle_{2^n} (2^{2n} - 1) \right. \\ \left. + \langle X_3 (2^{n-1} + 1) \right\rangle_{2^{n+1}} (2^n - 1) 2^n \right\rangle_{2^{3n} - 2^n}$$

③g

B. Computation of the mixed radix digits
for the attractive 3-moduli RNS.

Consider again the RNS based on the set
 $S = \{m_1, m_2, m_3\} = \{2^n - 1, 2^n, 2^n + 1\}$. Let
the RNS representation of the integer X be
 $X \xrightarrow{\text{RNS}} (X_1, X_2, X_3)$ where X_1, X_2, X_3 are
the residues of X .

The Mixed Radix formula is

$$X = X_1' + m_1 X_2' + m_1 m_2 X_3'$$

where X_1', X_2', X_3' are the mixed radix digits.

We have already shown that

$$X_1' = X_1 \quad (1)$$

$$X_2' = \langle m_1^{-1} (X_2 - X_1) \rangle_{m_2} \quad (2)$$

$$X_3' = \langle (m_1 m_2)^{-1} (X_3 - X_1 - m_1 X_2') \rangle_{m_3} \quad (3)$$

Considering now that $m_1 = 2^n - 1$, $m_2 = 2^n$, $m_3 = 2^n + 1$ one gets (4) g

$$\begin{aligned} X_2' &= \langle m_1^{-1} (X_2 - X_1) \rangle_{m_2} = \langle (2^n - 1)^{-1} (X_2 - X_1) \rangle_{2^n} \\ &= \langle (-1)^{-1} (X_2 - X_1) \rangle_{2^n} = \langle (-1) (X_2 - X_1) \rangle_{2^n} \\ &= \langle X_1 - X_2 \rangle_{2^n} \end{aligned}$$

or

$$\boxed{X_2' = \langle X_1 - X_2 \rangle_{2^n}} \quad (4)$$

The computation dictated by eq (4) is nothing more than $X_2' = X_1 + 2^n$'s complement of X_2 .

Regarding X_3' one gets from eq. (3)

$$\begin{aligned} X_3' &= \langle (m_1 m_2)^{-1} (X_3 - X_1 - m_1 X_2') \rangle_{m_3} \\ &= \langle [(2^n - 1) 2^n]^{-1} (X_3 - X_1 - (2^n - 1) X_2') \rangle_{2^{n+1}} \\ &= \langle [(-2)(-1)]^{-1} (X_3 - X_1 - (-1 - 1) X_2') \rangle_{2^{n+1}} \\ &= \langle 2^{-1} (X_3 - X_1 + 2 X_2') \rangle_{2^{n+1}} \\ &= \langle 2^{-1} (X_3 - X_1) + X_2' \rangle_{2^{n+1}} \end{aligned}$$

$$= \langle (2^{n-1} + 1)(X_3 - X_1) + X_2' \rangle_{2^{n+1}} \quad (5)g$$

(recall that $\langle 2^{-1} \rangle_{2^{n+1}} = \frac{2^n + 1}{2} = 2^{n-1} + 1$).

So finally

$$\boxed{X_3' = \langle (2^{n-1} + 1)(X_3 - X_1) + X_2' \rangle_{2^{n+1}}} \quad (5)$$

We'll now show how $\langle (2^{n-1} + 1)(X_3 - X_1) \rangle_{2^{n+1}}$ can be computed. Let S be

$S = \langle X_3 - X_1 \rangle_{2^{n+1}}$. Let the binary representation of S be $S = (s_n s_{n-1} \dots s_1 s_0)_2$

Then

$$\langle (2^{n-1} + 1)(X_3 - X_1) \rangle_{2^{n+1}} = \langle (2^{n-1} + 1)(s_n s_{n-1} \dots s_1 s_0)_2 \rangle_{2^{n+1}}$$

$$= \langle (2^{n-1} + 1) [(s_n s_{n-1} \dots s_1) \times 2 + s_0] \rangle_{2^{n+1}}$$

$$= \langle (2^n + 2)(s_n s_{n-1} \dots s_1) + (2^{n-1} + 1)s_0 \rangle_{2^{n+1}}$$

$$= \langle (s_n s_{n-1} \dots s_1) + (s_0 2^{n-1} + s_0) \rangle_{2^{n+1}}$$

$$= \langle (s_n s_{n-1} \dots s_1) + (s_0 \underbrace{000 \dots 000}_{n-2 \text{ zeros}} s_0) \rangle_{2^{n+1}}$$

$$= (s_n s_{n-1} \dots s_1) + (s_0 \underbrace{000 \dots 000}_{n-2 \text{ zeros}} s_0)$$

⑥ g

It will now be shown that

$$\begin{aligned} & \langle (s_n s_{n-1} \dots s_1) + (s_0 \underbrace{000 \dots 000}_{n-2 \text{ zeros}} s_0) \rangle_{2^{n+1}} \\ &= (s_n s_{n-1} \dots s_1)_2 + (s_0 \underbrace{000 \dots 000}_{n-2 \text{ zeros}} s_0) \end{aligned}$$

All that needs to be proven is that

$$(s_n s_{n-1} \dots s_1) + (s_0 000 \dots 000 s_0) < 2^{n+1}.$$

Since $S = (s_n s_{n-1} \dots s_1 s_0)_2 \in \mathbb{Z}_{2^{n+1}}$ (recall that

$$S = \langle X_3 - X_1 \rangle_{2^{n+1}}) \text{ then } S \in [0, 2^n].$$

If $s_n = 1$ then $s_{n-1} = s_{n-2} = \dots = s_1 = s_0 = 0$ and in this

$$\begin{aligned} \text{case } & (s_n s_{n-1} \dots s_1) + (s_0 000 \dots 000 s_0) = \\ &= (\underbrace{100 \dots 000}_{n-1 \text{ zeros}}) + (00 \dots 00) = 2^{n-1} < 2^{n+1}. \end{aligned}$$

If $s_n = 0$ then $(s_n s_{n-1} \dots s_1) + (s_0 \underbrace{000 \dots 000}_{n-2 \text{ zeros}} s_0) \leq$

$$\begin{aligned} & (\underbrace{0111 \dots 111}_{n-1 \text{ ones}}) + (\underbrace{1000 \dots 0001}_{n-2 \text{ zeros}}) = 2^{n-1} - 1 + 2^{n-1} + 1 \\ &= 2^n < 2^{n+1} \end{aligned}$$

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① h

Simplified arithmetic mod $(2^n + 1)$
using diminished-1 representations of
numbers

In the handout entitled "The 3-moduli RNS based on set $S = \{2^n - 1, 2^n, 2^n + 1\}$ " techniques for performing arithmetic mod $(2^n + 1)$ were offered; (see section B.3 starting on page 12).

A more efficient technique for performing arithmetic mod $(2^n + 1)$ is by using diminished-1 representations of numbers. This is the subject studied in this handout.

(2)h

[A]. Code translation

Let A be an integer such that $A \in \mathbb{Z}_{2^n+1}$ where $\mathbb{Z}_{2^n+1} = \{0, 1, 2, \dots, 2^n\}$; (the ring of integers mod (2^n+1)). Then A needs $n+1$ bits for its representation. The following shows the correspondence between normal and diminished-1 representations:

A	Diminished-1 of A or $A-1$
$0 = \overbrace{000 \dots 000}^{n+1 \text{ bit}}$	$\overbrace{100 \dots 000}^{n+1 \text{ bit}} = 2^n$
$1 = 000 \dots 001$	$00 \dots 000 = 0$
$2 = 000 \dots 010$	$00 \dots 001 = 1$
$3 = 000 \dots 011$	$00 \dots 010 = 2$
$4 = 000 \dots 100$	$00 \dots 011 = 3$
\vdots	
$2^n - 1 = 011 \dots 111$	11 $\dots 110 = 2^n - 2$
$2^n = 100 \dots 000$	$11 \dots 111 = 2^n - 1$

③ h

As seen from the previous correspondence, the diminished-1 representation of the number zero is $10\dots 0 = 2^n$ and requires $n+1$ bits for its representation. The diminished-1 representation of the non-zero elements $1, 2, \dots, 2^n-1, 2^n$ require only n bits for their representations.

B Arithmetic mod (2^n+1)

When performing arithmetic mod (2^n+1) by using the diminished-1 system, all input-operands are in diminished-1 form and the results are returned in diminished-1 form as well.

B.1 Addition mod (2^n+1)

Let A, B be integers such that $A, B \in \mathbb{Z}_{2^n+1}$. If $A=B=0$ then $\langle A+B \rangle_{2^n+1} = 0$;
If $A=0$ and $B \neq 0$ then $\langle A+B \rangle_{2^n+1} = B$;
If $A \neq 0$ and $B=0$ then $\langle A+B \rangle_{2^n+1} = A$.

(4)h

These cases (for which no addition is needed) can be detected and treated separately. The general case presented here is the case where $A \neq 0$, $B \neq 0$. The addition takes place as follows:

$$\begin{array}{l} A \xrightarrow{\text{dim}-1} A-1 \\ B \xrightarrow{\text{dim}-1} B-1 \end{array}$$

Add the n -bit numbers $A-1$ and $B-1$; complement (negate) the carry-out, end it around and add it back. The obtained result will be the diminished-1 form of $\langle A+B \rangle_{2^n+1}$.

Example 1: Perform $\langle A+B \rangle_{2^4+1}$ where

$$A = (10)_{10}, \quad B = (13)_{10}$$

$$A = 10 = (1010)_2 \xrightarrow{\text{dim}-1} (1001)_2 = A-1$$

$$B = 13 = (1101)_2 \xrightarrow{\text{dim}-1} (1100)_2 = B-1$$

Adding $A-1$ and $B-1$ we get

$$\begin{array}{r} 1001 \\ + 1100 \\ \hline 10101 \\ \xrightarrow{\text{around}} \\ 0101 = 5 = \text{dim}-1 \text{ of } 6. \end{array}$$

Double check to see that $\langle A+B \rangle_{2^4+1} \stackrel{\textcircled{5}h}{=} \langle 10+13 \rangle_{17} = 6$ and what we got is the diminished-1 form of 6.

Example 2: Perform $\langle A+B \rangle_{2^4+1}$ where

$$A = (8)_{10}, \quad B = (7)_{10}.$$

$$A = 8 = (1000)_2 \xrightarrow{\text{dim-1}} 0111 = A-1$$

$$B = 7 = (0111)_2 \xrightarrow{\text{dim-1}} 0110 = B-1$$

Adding $A-1$ and $B-1$ yields

$$\begin{array}{r} 0111 \\ 0110 \\ \hline 01101 \\ \hline 1110 = 14 = \text{dim-1 of } 15. \end{array}$$

Double check to see that $\langle 8+7 \rangle_{17} = 15$ and what we got is the dim-1 form of 15.

Example 3: Perform $\langle A+B \rangle_{2^4+1}$ where

$$A = (8)_{10}, \quad B = (9)_{10}$$

⑥ h

Here $A = (8)_{10} = 1000 \xrightarrow{\text{dim-1}} 0111 = A-1$
 $B = (9)_{10} = 1001 \xrightarrow{\text{dim-1}} 1000 = B-1$

$$A-1 + B-1 = \begin{array}{r} 0111 \\ 1000 \\ \hline 0111 \\ \hline 1000 \end{array} = 2^4 = \text{dim-1 form of zero.}$$

See that $\langle 8+9 \rangle_{17} = 0$

B.2 Additive inverse mod (2^n+1) ; (negation mod (2^n+1))

Let A belong to Z_{2^n+1} . If $A=0$ then $\langle -A \rangle_{2^n+1} = 0$; (no computation required; this case can be detected and treated separately). For the general case where $A \neq 0$ the computation $\langle -A \rangle_{2^n+1}$ occurs as follows:

$$A \xrightarrow{\text{dim-1}} A-1$$

Take the 1's complement of $A-1$. The obtained result will be the diminished-1 form of $\langle -A \rangle_{2^n+1}$.

Example 4: Compute $\langle -A \rangle_{2^4+1}$ where $\textcircled{7}h$

$$A = (10)_{10}.$$

$$\text{Here } A = (10)_{10} = (1010)_2 \xrightarrow{\text{dim-1}} (1001)_2 = A-1.$$

$$\overline{A-1} = (0110)_2 = 6 = \text{dim-1 form of } 7.$$

Double check to see that $\langle -10 \rangle_{17} = 7$
and what we got is its dim-1 form.

B.3 Subtraction mod (2^n+1)

Obviously subtraction can be accommodated by a negation and an addition

B.4 Scaling by power of two (scaling by 2^k)

Let A belong to \mathbb{Z}_{2^n+1} . If $A=0$ then

$$\langle 2^k A \rangle_{2^n+1} = 0; \text{ (no computation required;}$$

this case can be detected and treated separately). For the general case where

$A \neq 0$ the computation $\langle 2^k A \rangle_{2^n+1}$ takes

place as follows:

⑧ h

$$A \xrightarrow{\text{dim}-1} A-1$$

Rotate the number $A-1$ k bits to the left where the bits ~~shifted out of the left end~~ shifted out of the left end and shifted into the right end must be complemented (negated)

The obtained result will be the diminished-1 form of $\langle 2^k A \rangle_{2^{n+1}}$

Example 5: Compute $\langle A \times 2^3 \rangle_{2^4+1}$ where

$$A = (10)_{10}$$

$$\text{Here } A=10 = (1010)_2 \xrightarrow{\text{dim}-1} (1001)_2 = A-1.$$

$$1001 \xrightarrow[\text{etc}]{\substack{\text{rot 3-bit} \\ \text{left} \\ \text{compl.}}} 1\bar{1}\bar{0}\bar{0} = (1011)_2 = 11 =$$

= diminished-1 form of 12.

Double check to see that $\langle A \times 2^3 \rangle_{2^4+1} =$

$$= \langle 10 \times 8 \rangle_{17} = 12 \text{ and what we got is its}$$

dim-1 form.

9h

Example 6:

$$\text{Compute } \langle 2^3(A-B) + 2^4(C-D) \rangle_{2^{5+1}}$$

where $A=11, B=6, C=13, D=9$.

Solution:

$$A = (11)_{10} = (01011)_2 \xrightarrow{\text{dim}-1} A_{-1} = \text{0010} (01010)_2$$

$$B = (6)_{10} = (00110)_2 \xrightarrow{\text{dim}-1} B_{-1} = (00101)_2$$

$$C = (13)_{10} = (01101)_2 \xrightarrow{\text{dim}-1} C_{-1} = (01100)_2$$

$$D = (9)_{10} = (01001)_2 \xrightarrow{\text{dim}-1} D_{-1} = (01000)_2$$

The additive inverses of B and D in diminished-1 form are

$$\overline{B_{-1}} = (11010)_2 = 1's \text{ compl. of } B_{-1}$$

$$\overline{D_{-1}} = (10111)_2 = \text{ " " " } D_{-1}$$

$$\text{dim}-1 \text{ of } \langle A-B \rangle_{2^{5+1}} = \begin{array}{r} 01010 \\ +) 11010 \\ \hline 100100 \\ \hline 00100 \end{array}$$

(10) h

dim-1 of $\langle 2^3(A-B) \rangle_{2^5+1}$ is

$$00100 \xrightarrow[\text{compl. etc.}]{\text{rotate 3-bct left}} 00\bar{0}\bar{0}\bar{1} = 00110$$

$$\text{dim-1 of } \langle C-D \rangle_{2^5+1} = \begin{array}{r} 01100 \\ +) 10111 \\ \hline 100011 \\ \hline 0 \\ \hline 00011 \end{array}$$

dim-1 of $\langle 2^4(C-D) \rangle_{2^5+1}$ is

$$00011 \xrightarrow[\text{compl. etc.}]{\text{rot. 4-bct left}} 1\bar{0}\bar{0}\bar{0}\bar{1} = 11110$$

Finally

dim-1 form of $\langle 2^3(A-B) + 2^4(C-D) \rangle_{2^5+1}$

$$= \begin{array}{r} 00110 \\ +) 11110 \\ \hline 100100 \\ \hline 0 \\ \hline \end{array}$$

$$(00100)_2 = 4 = \text{dim-1 form of 5.}$$

⑪ h

Double check to see that

$$\begin{aligned} & \langle 2^3(A-B) + 2^4(C-D) \rangle_{2^5+1} = \\ & = \langle 2^3(11-6) + 2^4(13-9) \rangle_{33} = \langle 8 \times 5 + 16 \times 4 \rangle_{33} = \\ & = \langle 40 + 64 \rangle_{33} = \langle 104 \rangle_{33} = 5 \text{ and what we got is} \\ & (00100) \text{ which is dim-1 of 5.} \end{aligned}$$

- This composite example makes it clear that in composite calculations, after entering the diminished-1 system we stay in the diminished-1 system all the time until the final result is computed. For this particular example, the inputs A, B, C, D were translated into their diminished-1 forms, and after that we stayed inside the diminished-1 system because all the intermediate results were in diminished-1 form.

(12)h

B.5 Multiperand addition mod (2^{n+1})

Let $A_1, A_2, A_3, \dots, A_k$ be integers belonging to $\mathbb{Z}_{2^{n+1}}^n$. We are interested in computing

$\langle A_1 + A_2 + \dots + A_k \rangle_{2^{n+1}}$ using the dim-1

approach. If one or more numbers A_i are zero, they do not contribute to the summation. For the general case where $A_i \neq 0, \forall i=1, 2, \dots, k$ the computation

$\langle \sum_{i=1}^k A_i \rangle_{2^{n+1}}$ takes place as follows:

$$A_1 \xrightarrow{\text{dim-1}} A_1 - 1$$

$$A_2 \xrightarrow{\text{dim-1}} A_2 - 1$$

$$A_3 \xrightarrow{\text{dim-1}} A_3 - 1$$

$$\vdots \quad \quad \quad \vdots$$

$$A_k \xrightarrow{\text{dim-1}} A_k - 1$$

Add the k diminished-1 forms $A_1 - 1,$

$A_2 - 1, A_3 - 1, \dots, A_k - 1$ using a CSA tree

followed by a CPA (a 2-operand adder). The CSA tree will be a minimum delay (minimum # of levels) tree and will reduce the k numbers (rows) down to two. The CPA will add the two vectors to produce the final result. The CSA tree will consist of $(2,2)$, $(3,2)$ counters or in general counters of arbitrary size (say (p,q) counters). Both the CSA tree as well as the CPA will use the scheme complement the carry-out and end it around.

The obtained result will be the $\text{dim}-1$ form of $\langle A_1 + A_2 + \dots + A_k \rangle_{2^{n+1}}$.

Example 7: Compute $\langle A+B+C+D \rangle_{2^4+1}$ where

$$A = 12, B = 11, C = 10, D = 8.$$

Here

$$A = 12 = (1100)_2 \xrightarrow{\text{dim}-1} A_{-1} = (1011)_2$$

$$B = 11 = (1011)_2 \xrightarrow{\text{dim}-1} B_{-1} = (1010)_2$$

(14) h

$$C = 10 = (1010)_2 \xrightarrow{\text{dim}-1} C-1 = (1001)_2$$

$$D = 8 = (1000)_2 \xrightarrow{\text{dim}-1} D-1 = (0111)_2$$

The counters used here will be (3,2) counters. The sequence of #s is 3, 4, 6, 9, ... and obviously two levels of reduction will be needed to perform the 4-to-2 reduction. Then a CPA will give the final result.

$$\begin{array}{r}
 A-1 = \begin{array}{|c|} \hline 1 \\ \hline \end{array} \begin{array}{|c|} \hline 0 \\ \hline \end{array} \begin{array}{|c|} \hline 1 \\ \hline \end{array} \begin{array}{|c|} \hline 1 \\ \hline \end{array} \\
 B-1 = \begin{array}{|c|} \hline 1 \\ \hline \end{array} \begin{array}{|c|} \hline 0 \\ \hline \end{array} \begin{array}{|c|} \hline 1 \\ \hline \end{array} \begin{array}{|c|} \hline 0 \\ \hline \end{array} \\
 C-1 = \begin{array}{|c|} \hline 1 \\ \hline \end{array} \begin{array}{|c|} \hline 0 \\ \hline \end{array} \begin{array}{|c|} \hline 0 \\ \hline \end{array} \begin{array}{|c|} \hline 1 \\ \hline \end{array} \\
 D-1 = 0 \quad 1 \quad 1 \quad 1
 \end{array}$$

$$\begin{array}{|c|} \hline 0 \\ \hline \end{array} \begin{array}{|c|} \hline 1 \\ \hline \end{array} \begin{array}{|c|} \hline 1 \\ \hline \end{array} \begin{array}{|c|} \hline 1 \\ \hline \end{array} \\
 \begin{array}{|c|} \hline 0 \\ \hline \end{array} \begin{array}{|c|} \hline 1 \\ \hline \end{array} \begin{array}{|c|} \hline 1 \\ \hline \end{array} \begin{array}{|c|} \hline 0 \\ \hline \end{array} \\
 \begin{array}{|c|} \hline 1 \\ \hline \end{array} \begin{array}{|c|} \hline 0 \\ \hline \end{array} \begin{array}{|c|} \hline 0 \\ \hline \end{array} \begin{array}{|c|} \hline 1 \\ \hline \end{array}$$

$$\begin{array}{cccc}
 1 & 1 & 0 & 1 \\
 1 & 0 & 0 & 0
 \end{array}$$

CPA

$$\begin{array}{cccc}
 1 & 0 & 1 & 1 & 0 \\
 \hline
 & & & & 0
 \end{array}$$

$$0110 = 6 = \text{dim}-1 \text{ of } 7$$

Double check to see that

$\langle A+B+C+D \rangle_{2^4+1} = \langle 12+11+10+8 \rangle_{17} = 7$ and what $(15)_h$ we got is its dim-1 form.

B.6 General multiplication mod (2^n+1)

An efficient multiplier-design for performing $\langle A \times B \rangle_{2^n+1}$ using the diminished-1 approach is offered in reference [2]. The interested reader can find more details on arithmetic mod (2^n+1) using the dim-1 approach in references [1], [2]

References:

- [1] L.M. Leibowitz, "A simplified binary arithmetic for the Fermat number transform", IEEE Trans. Acoustics, Speech, and Signal Processing, vol. ASSP-24, no. 5, pp. 356-359, Oct. 1976.
- [2] Z. Wang, G.A. Jullien, and W.C. Miller, "An algorithm for multiplication modulo (2^n+1) ", in Proceedings of 29th Asilomar Conference on Signals, Systems, and Computers, (Pacific Grove, CA, Oct. 1995), pp. 956-960.